



# Review Realistic Example

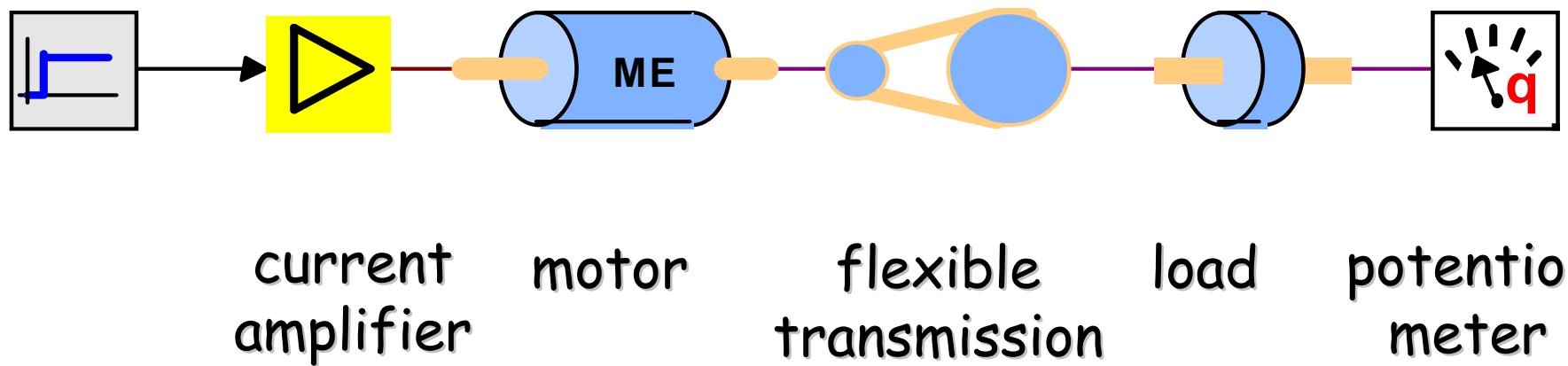
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University of Twente, Netherlands

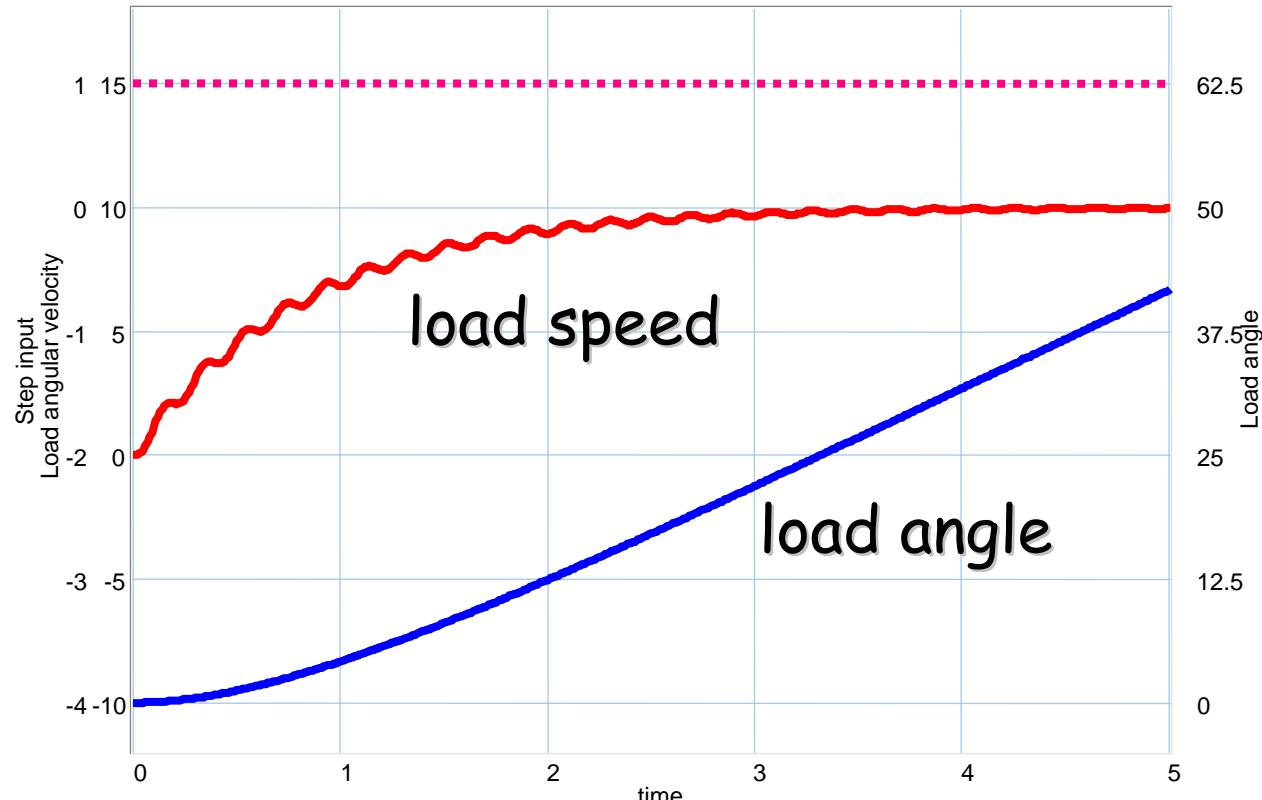
[www.ce.utwente.nl/amn](http://www.ce.utwente.nl/amn)  
[J.vanAmerongen@utwente.nl](mailto:J.vanAmerongen@utwente.nl)

- Design of controllers for a realistic servo system
- Fourth order servo system
  - Tacho feedback
  - Saturation

# Example of lecture 2

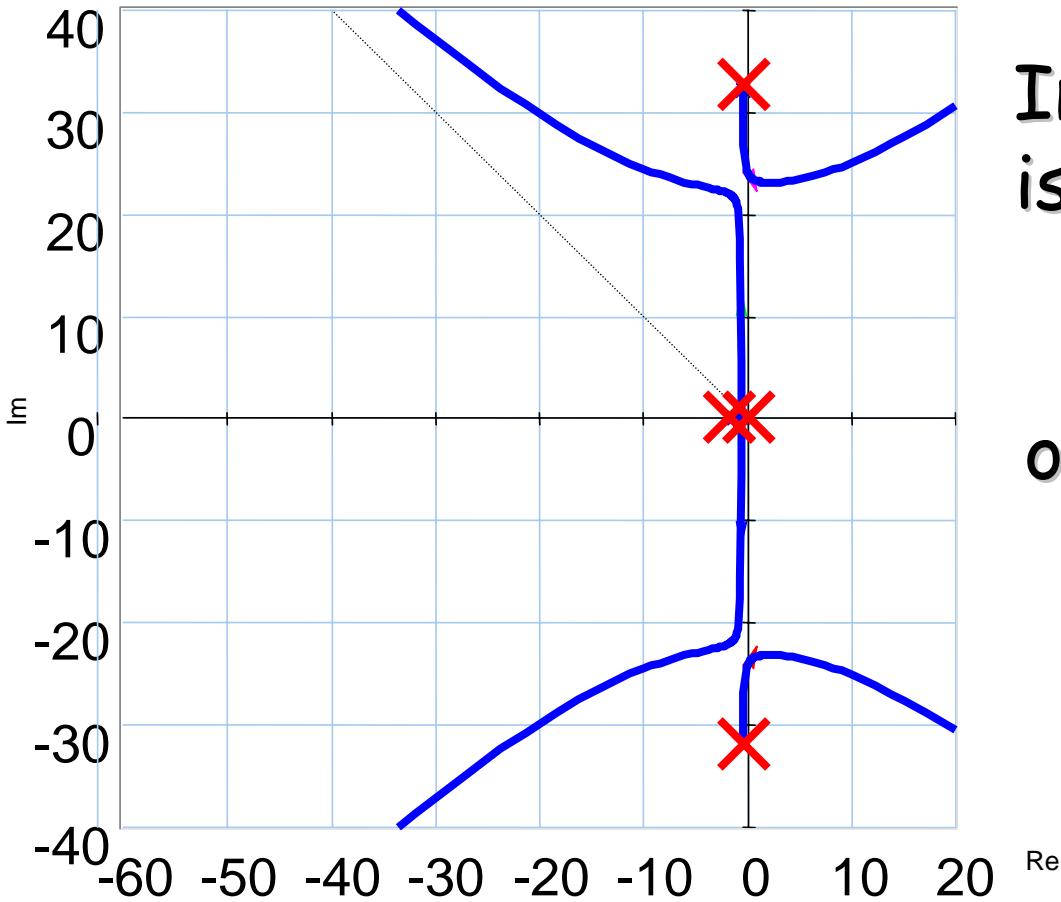


# Example of lecture 2



A\_Model\_for\_identification.em

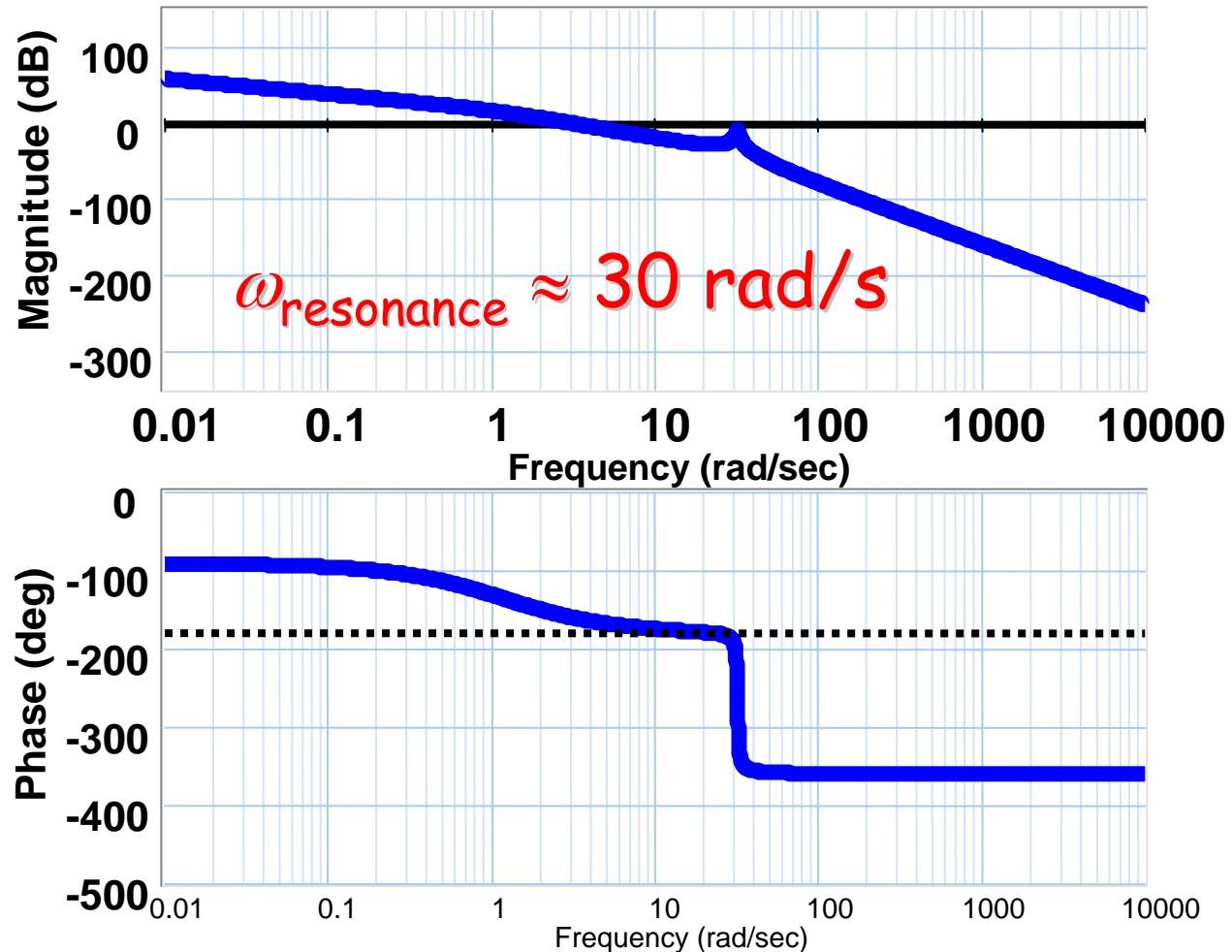
# Linearise in 20-sim



In practice this option is not always available:

Black Box modelling  
or Grey Box modelling

# Bode plot

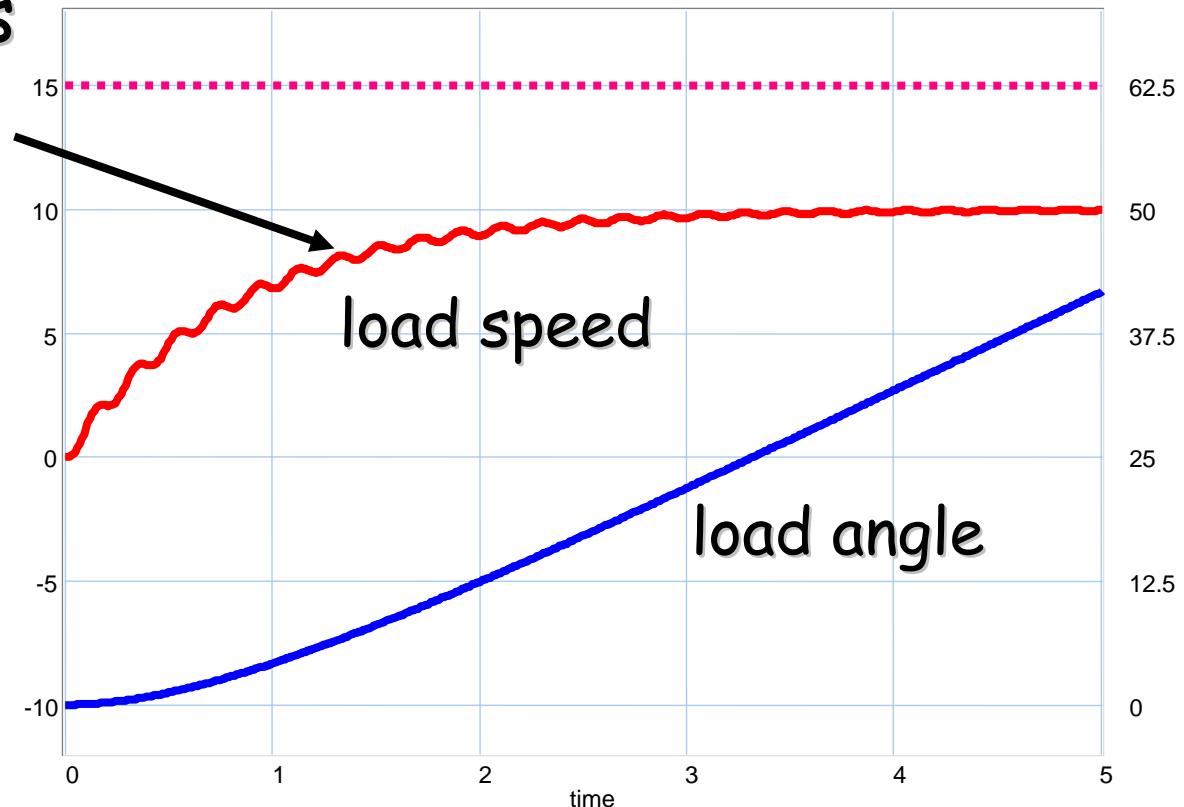


# Example of lecture 2

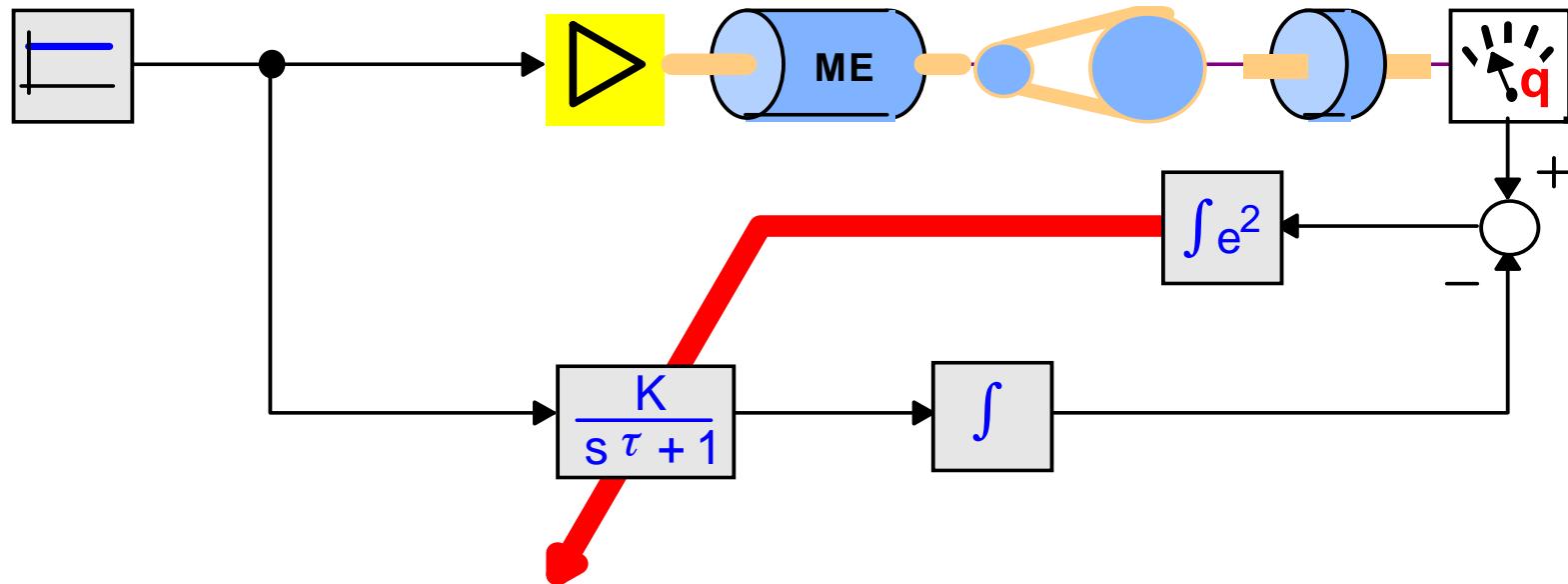
$T_{\text{resonance}} \approx 0.2 \text{ s}$

$\omega_{\text{resonance}} \approx 30$

Response of  
load angle  
suggests poles  
in 0 and - $\alpha$



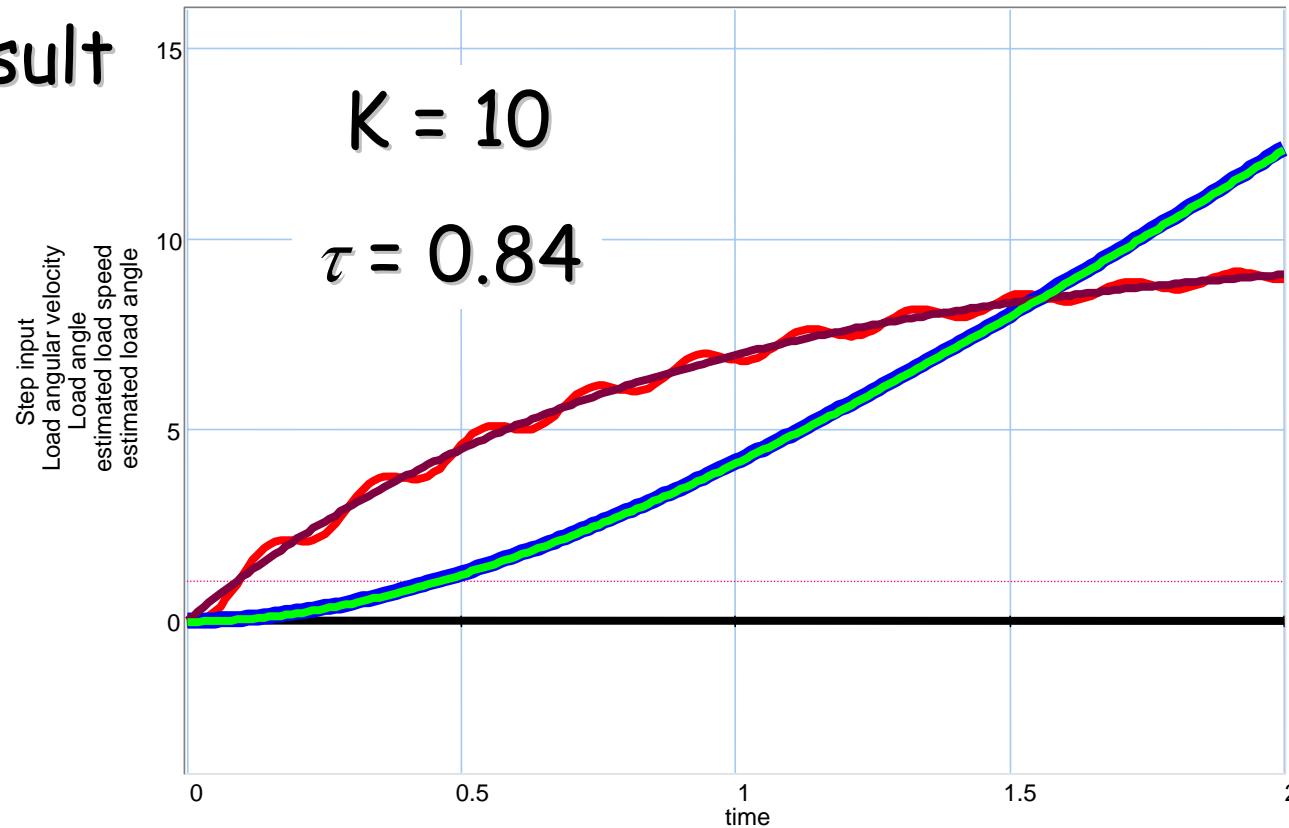
# Identification experiment



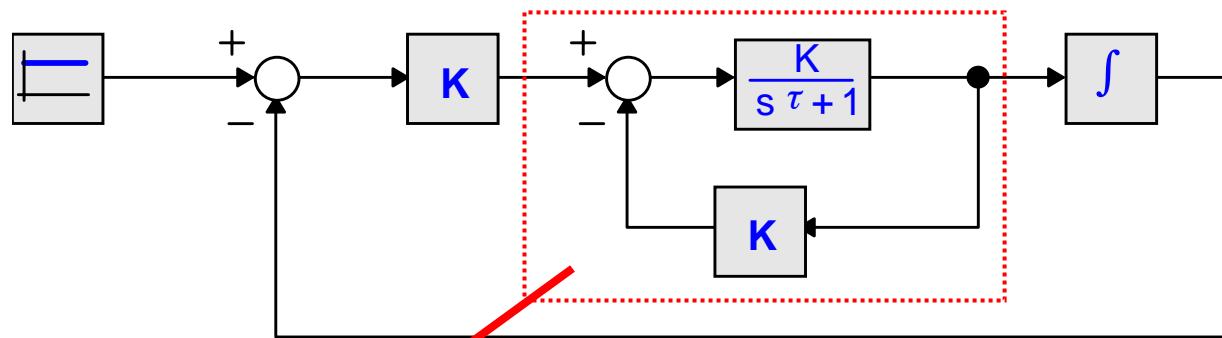
A1\_Model\_for\_(exp)\_identification.em

# Identification experiment

Result



# Proposed control structure



$$\frac{K'}{s + a + K'K_d}$$

$$\frac{K'K_p}{s(s + a + K'K_d) + K'K_p}$$

$$-\frac{1}{K_d} = \frac{K's}{s(s + a) + K'K_p}$$

$$-\frac{1}{K_p} = \frac{K'}{s(s + a)}$$

# Root locus $K_p$

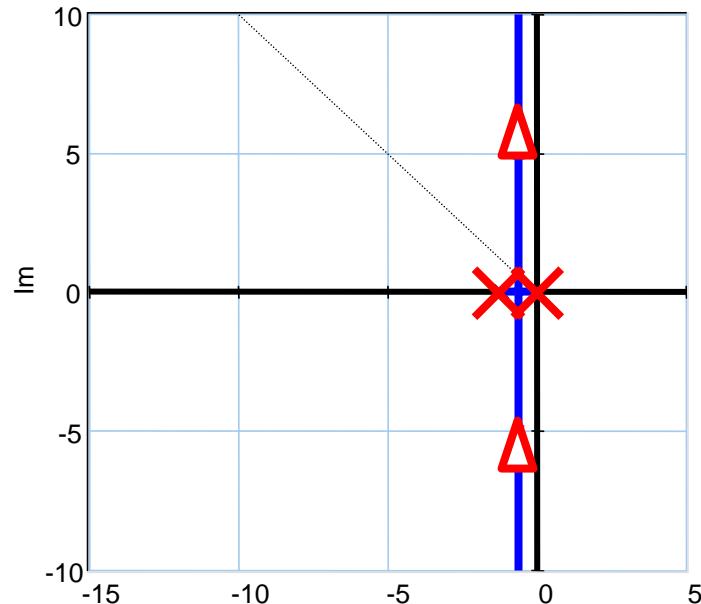
$$-\frac{1}{K_p} = \frac{K'}{s(s+a)}$$

$$a = -1.2 \\ K' = 10 \cdot 1.2 = 12$$

Stay away from the resonance frequency (30 rad/s):

e.g.  $\omega_n = 6$  rad/s

$$K_p \approx 36/12 = 3$$

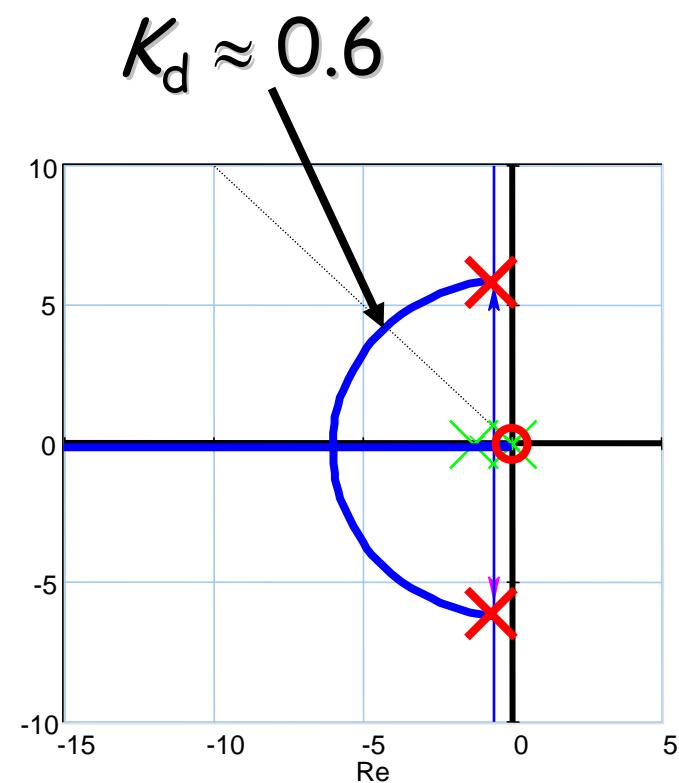


Design goal based on mechanical limitations

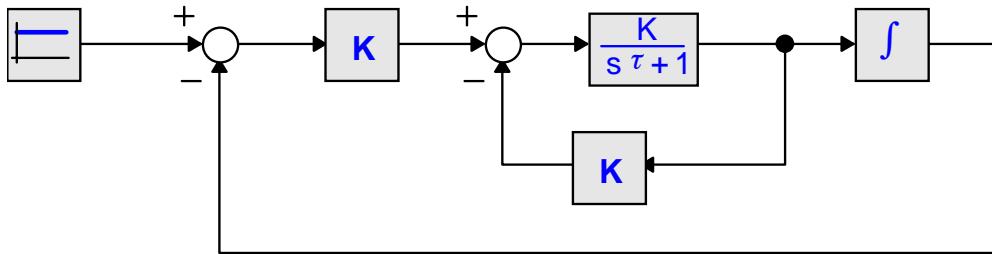
# Root locus $K_d$

$$-\frac{1}{K_d} = \frac{K's}{s(s+a) + K'K_p}$$

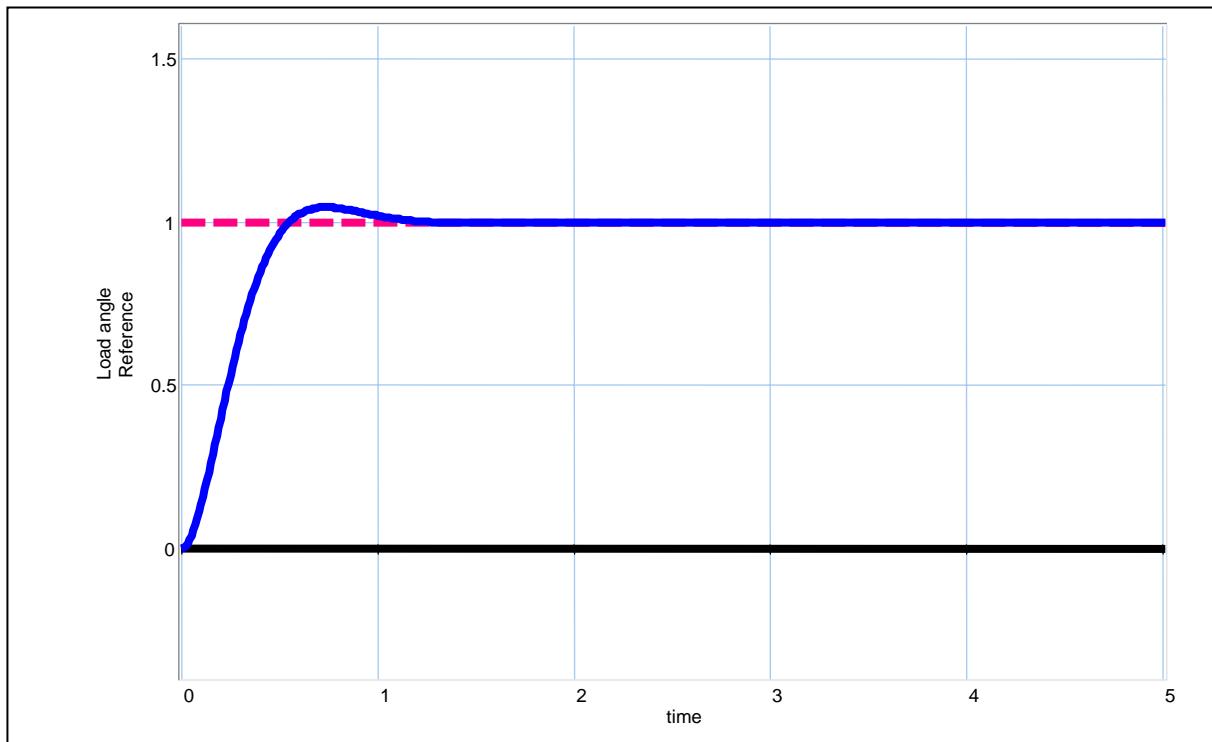
In a similar way  
we can find  
for  $K_p = 4$   
 $K_d = 0.8$   
( $\omega_n = 7$  rad/s)



# Response of 'ideal' system



$$K_p = 3$$
$$K_d = 0.6$$



A2\_Prop\_plus\_tacho\_FB\_exp\_model.em

# Alternative design

- Use state space representation and select  $\omega_n$  (factor 5 smaller than resonance frequency) and desired  $z$

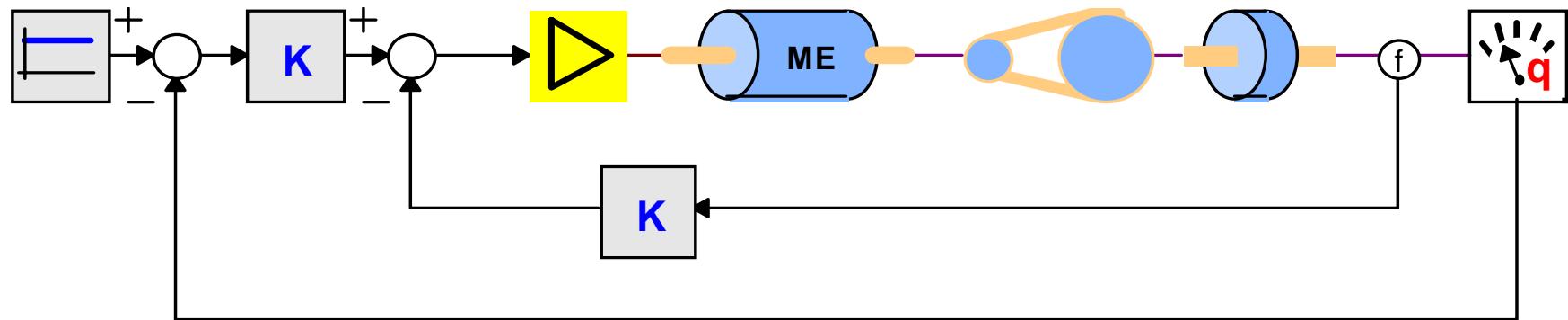
$$A' = \begin{pmatrix} 0 & 1 \\ -K'K_p & -a - K'K_d \end{pmatrix} = \begin{pmatrix} 0 & 1 \\ -\omega_n^2 & -2z\omega_n \end{pmatrix}$$

$$A' = \begin{pmatrix} 0 & 1 \\ -12K_p & -1.2 - 12K_d \end{pmatrix} = \begin{pmatrix} 0 & 1 \\ -36 & -8.4 \end{pmatrix}$$

$$K_p = 3$$

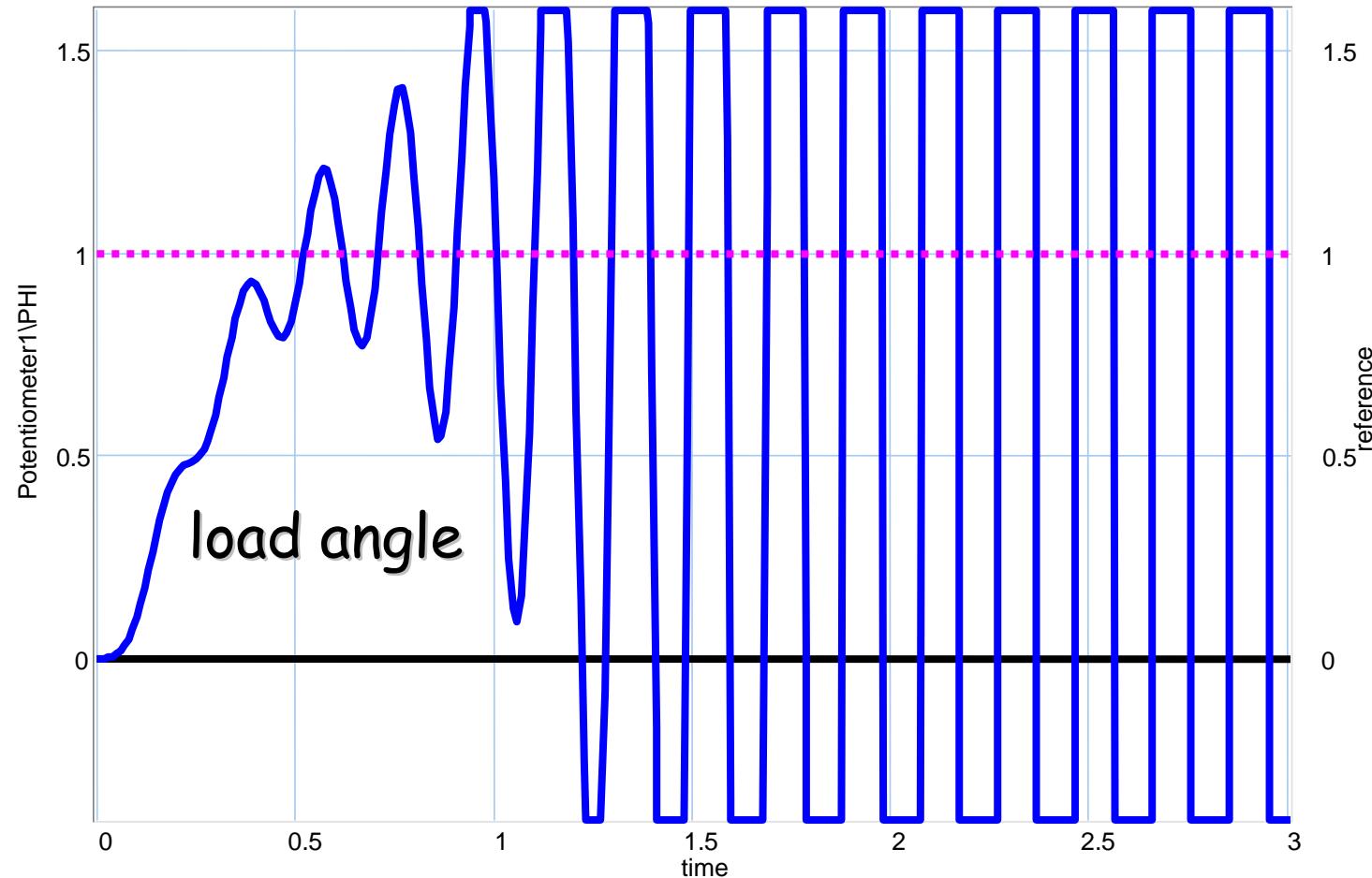
$$K_d = 0.6$$

# In “real” system

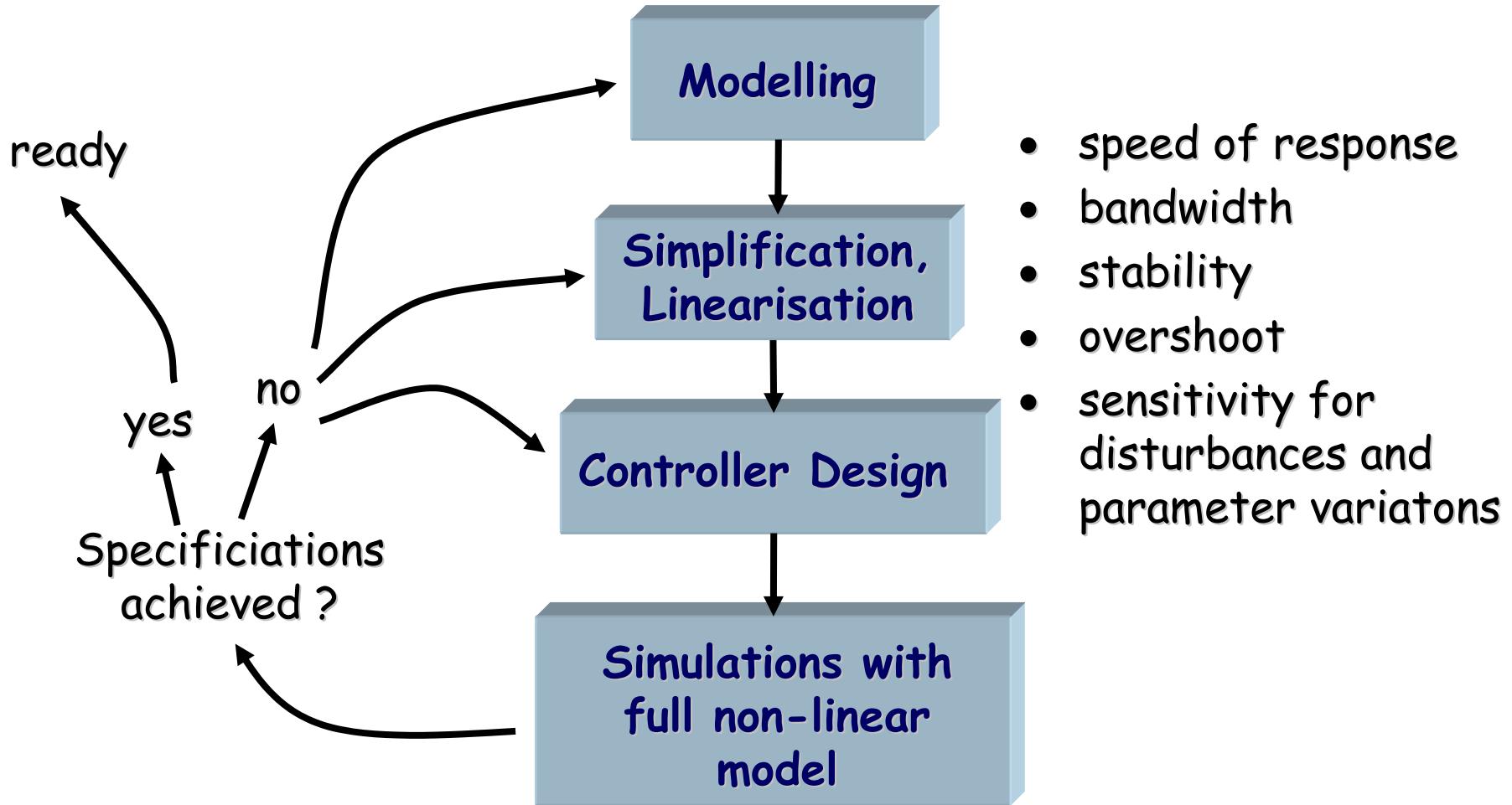


A3\_Prop\_plus\_tacho\_FB\_real\_model.em

# Response of 'real' system



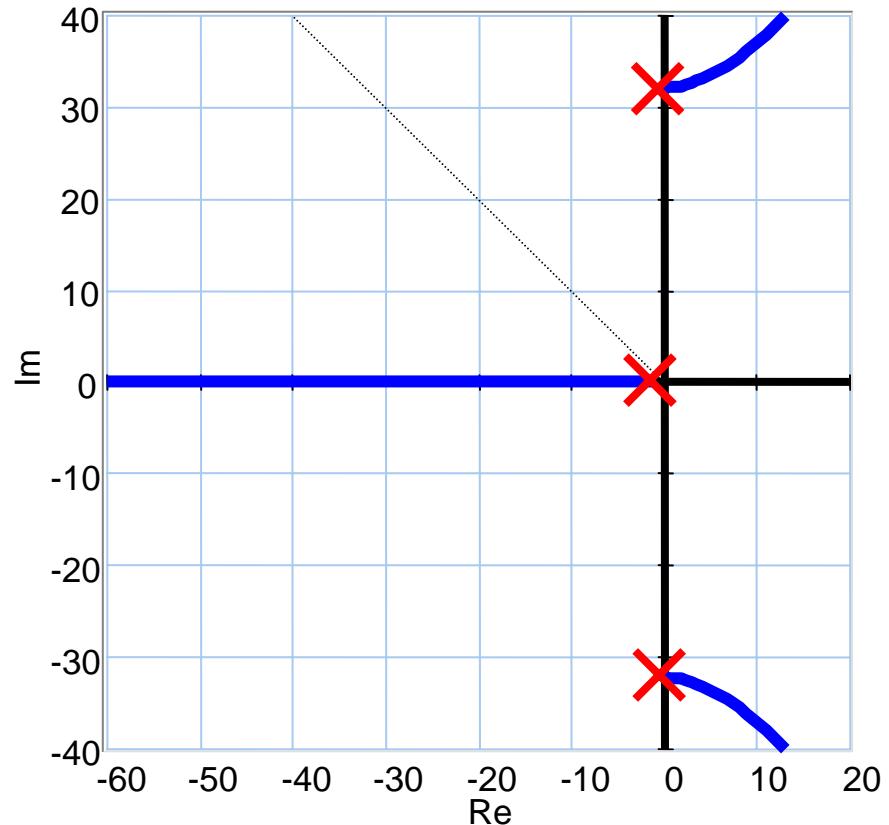
# Design issues (from lecture 1)



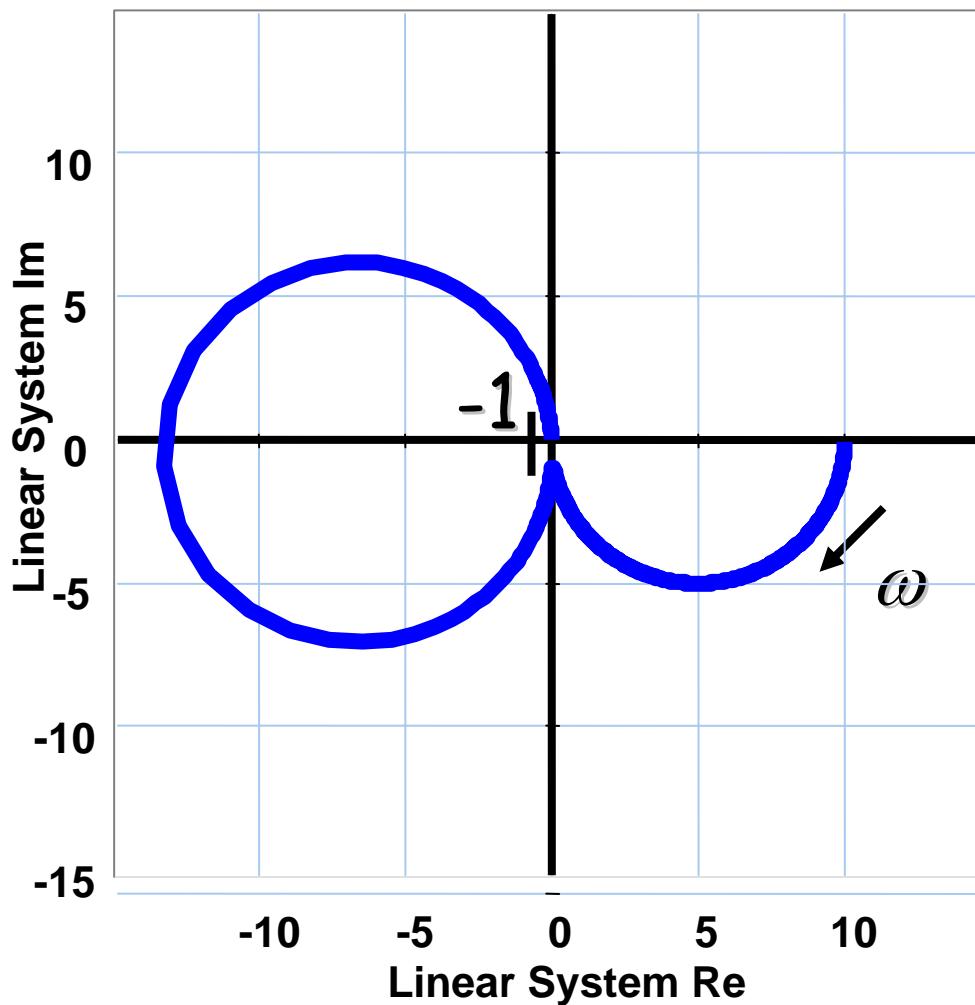
# Tacho feedback in “real” system

A\_Model\_for\_identification.em  
(tacho -feedback)

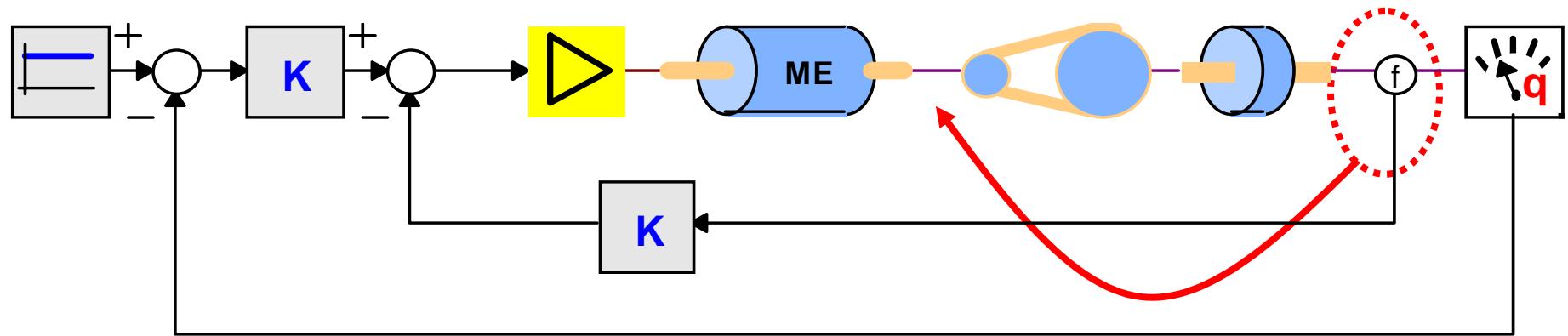
Due to  
(disregarded)  
complex poles,  
tacho feedback  
of no use here



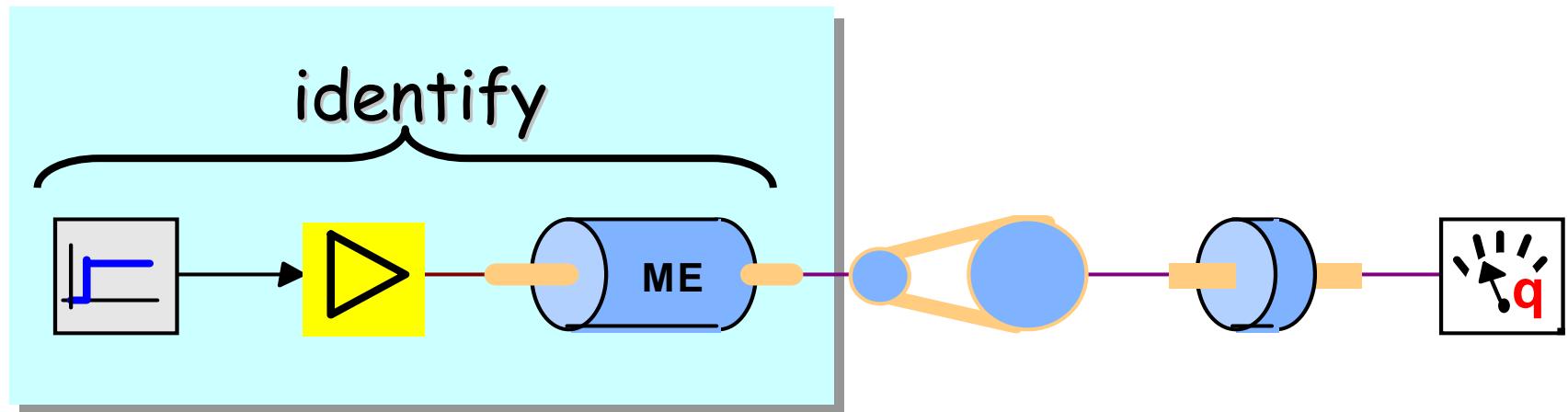
# Nyquist



# Alternative tacho feedback



# New linearisation

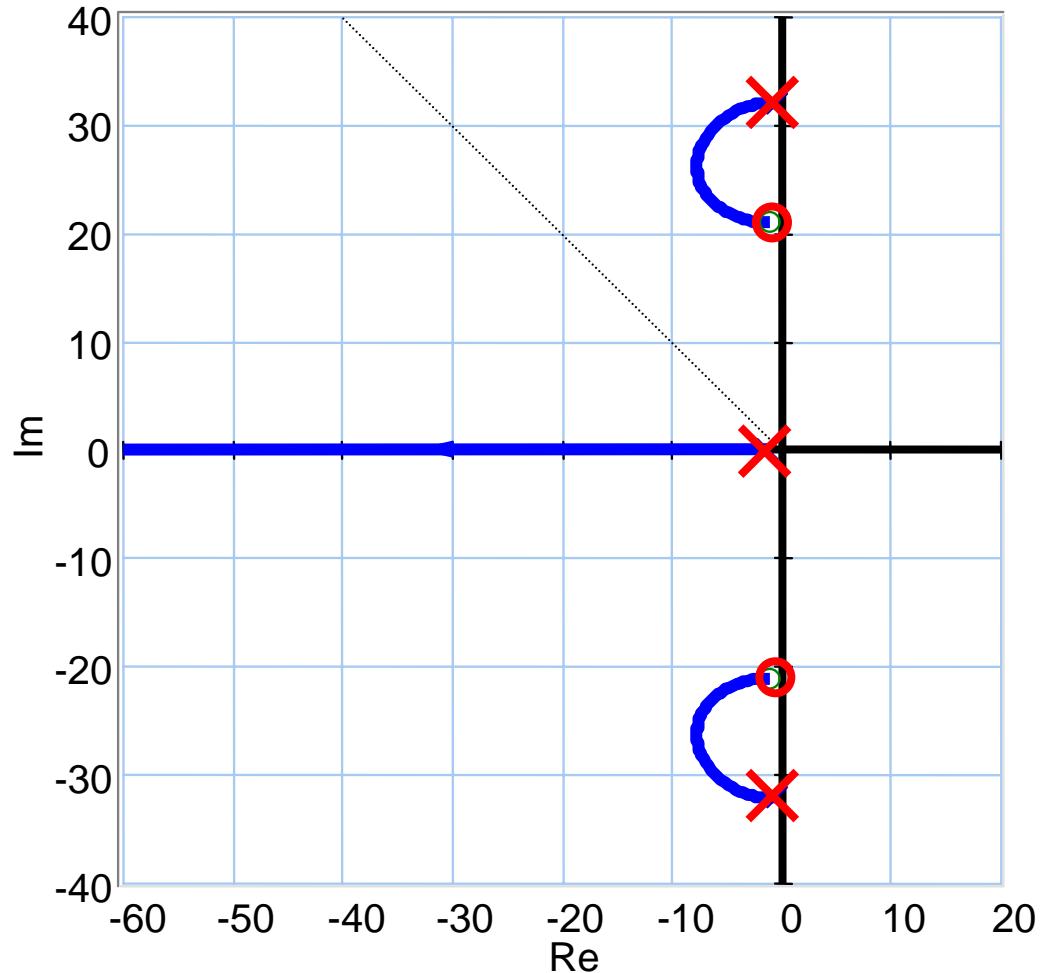


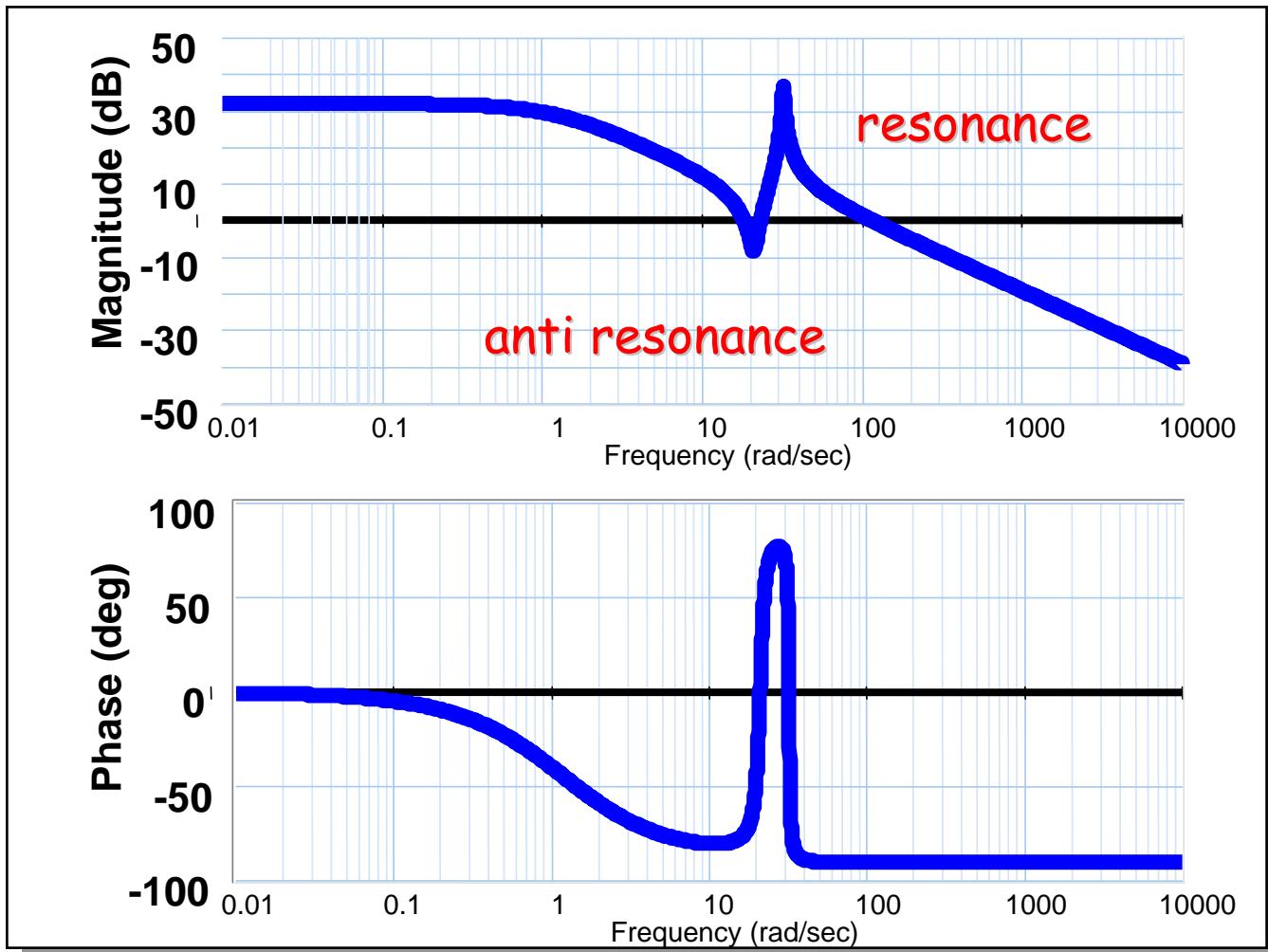
A\_Model\_for\_identification.em

# This helps !

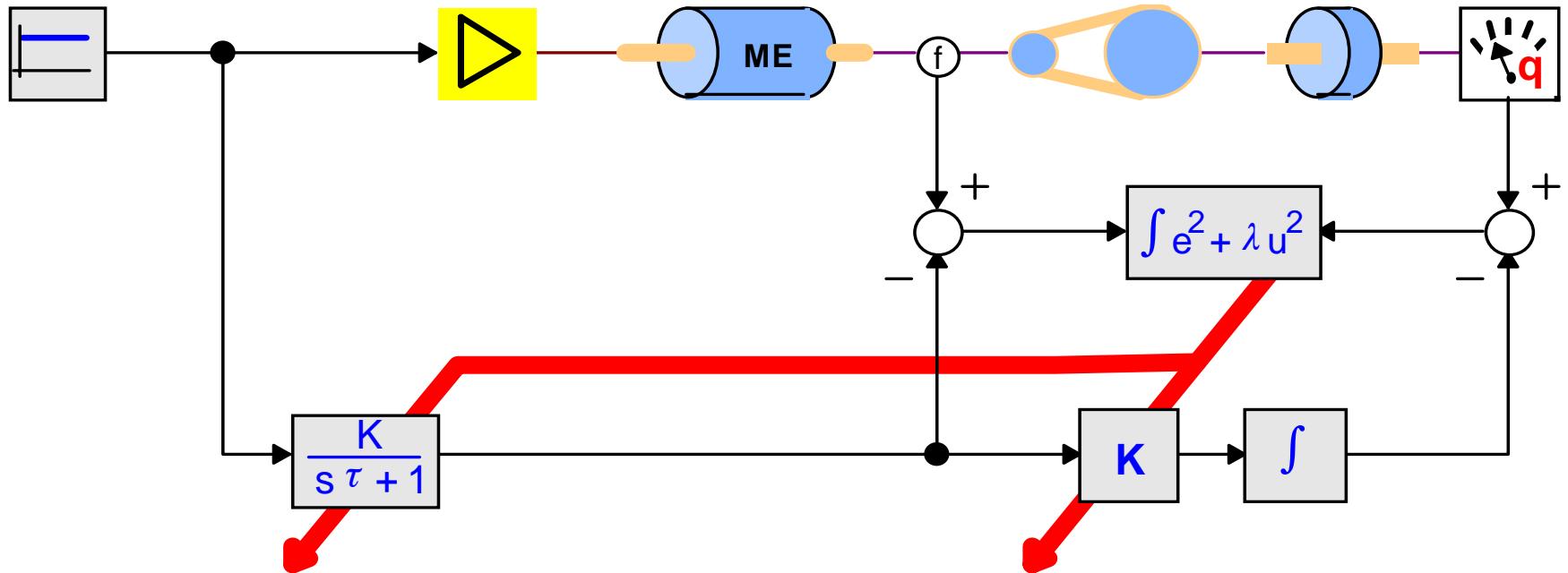
Complex poles  
now compensated  
by complex zeros

On real axis  
tacho feedback  
does what it is  
expected to do !



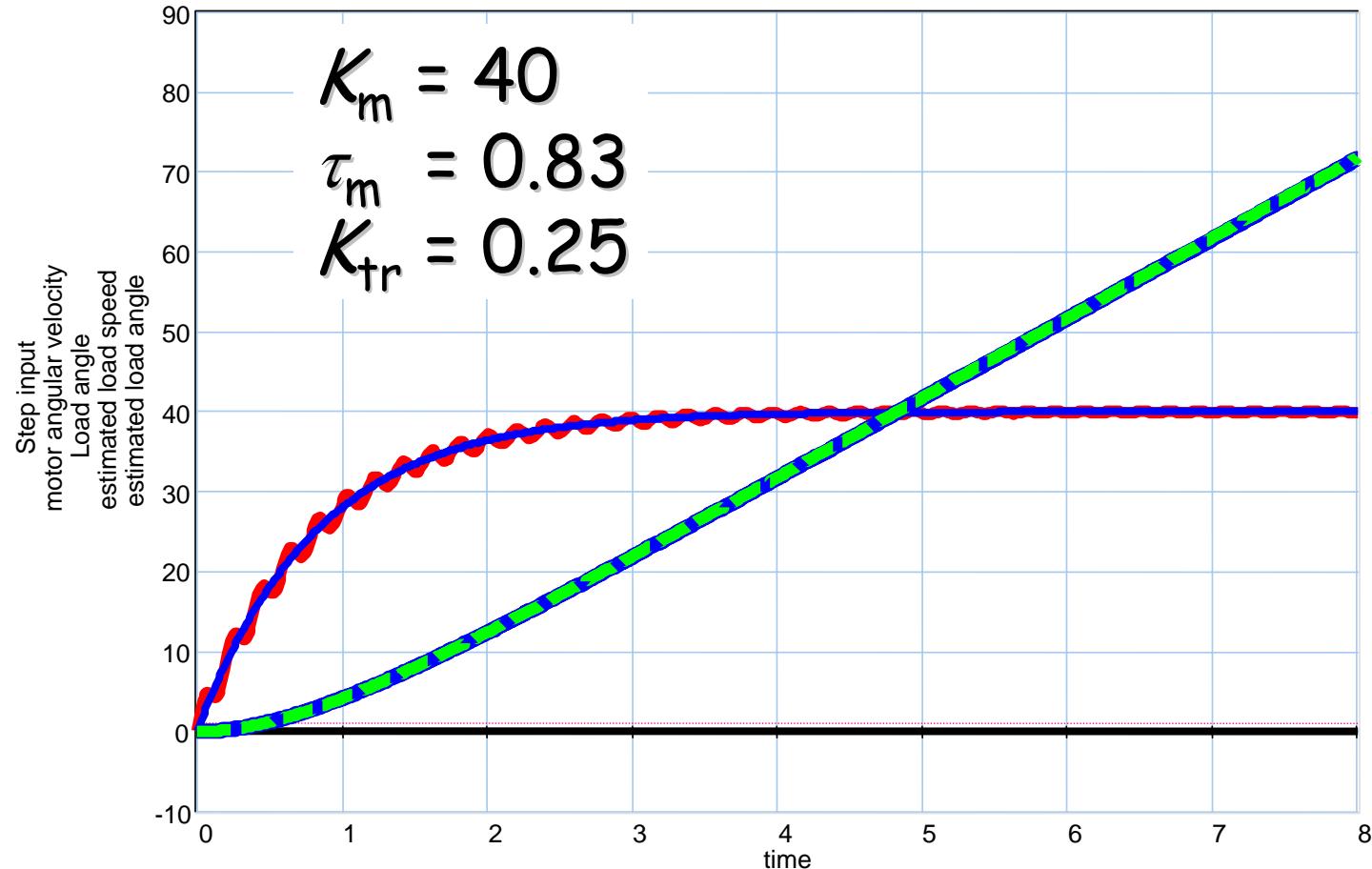


# Experiment



A4\_Model\_for\_(exp)\_identification\_motor\_tacho\_FB.em

# Identification result



We found before:

$$K_p = 3$$

$$K_d = 0.6$$

Because gain from  
input to motor is  
4 times higher,

or

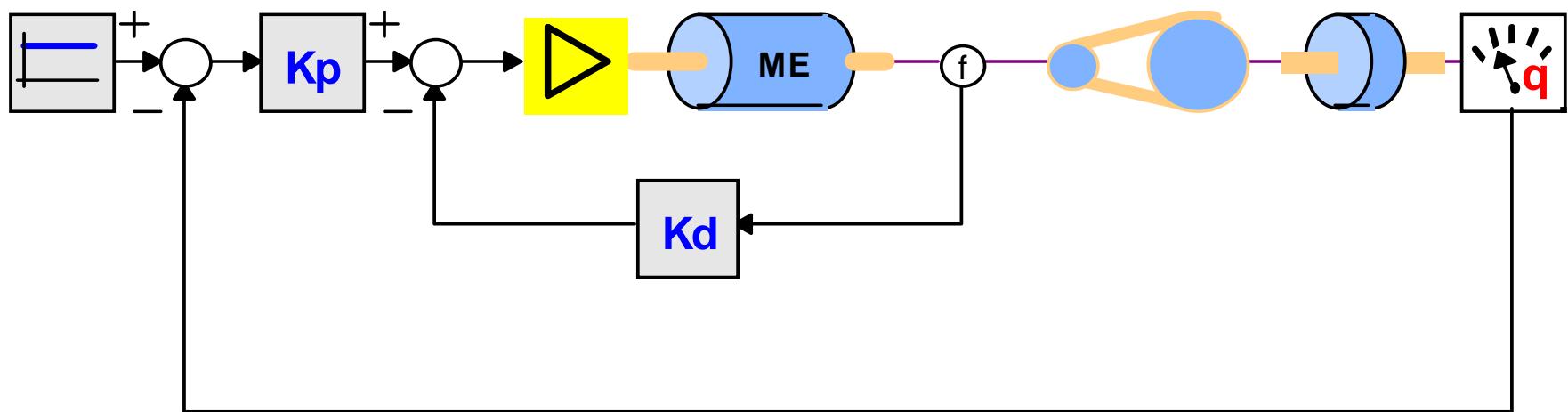
$$K_{d,m} = 0.6/4 = 0.15 \quad (0.2)$$

$$K_p = 4$$

$$K_d = 0.8$$

$$K_p = 3 \quad (4)$$

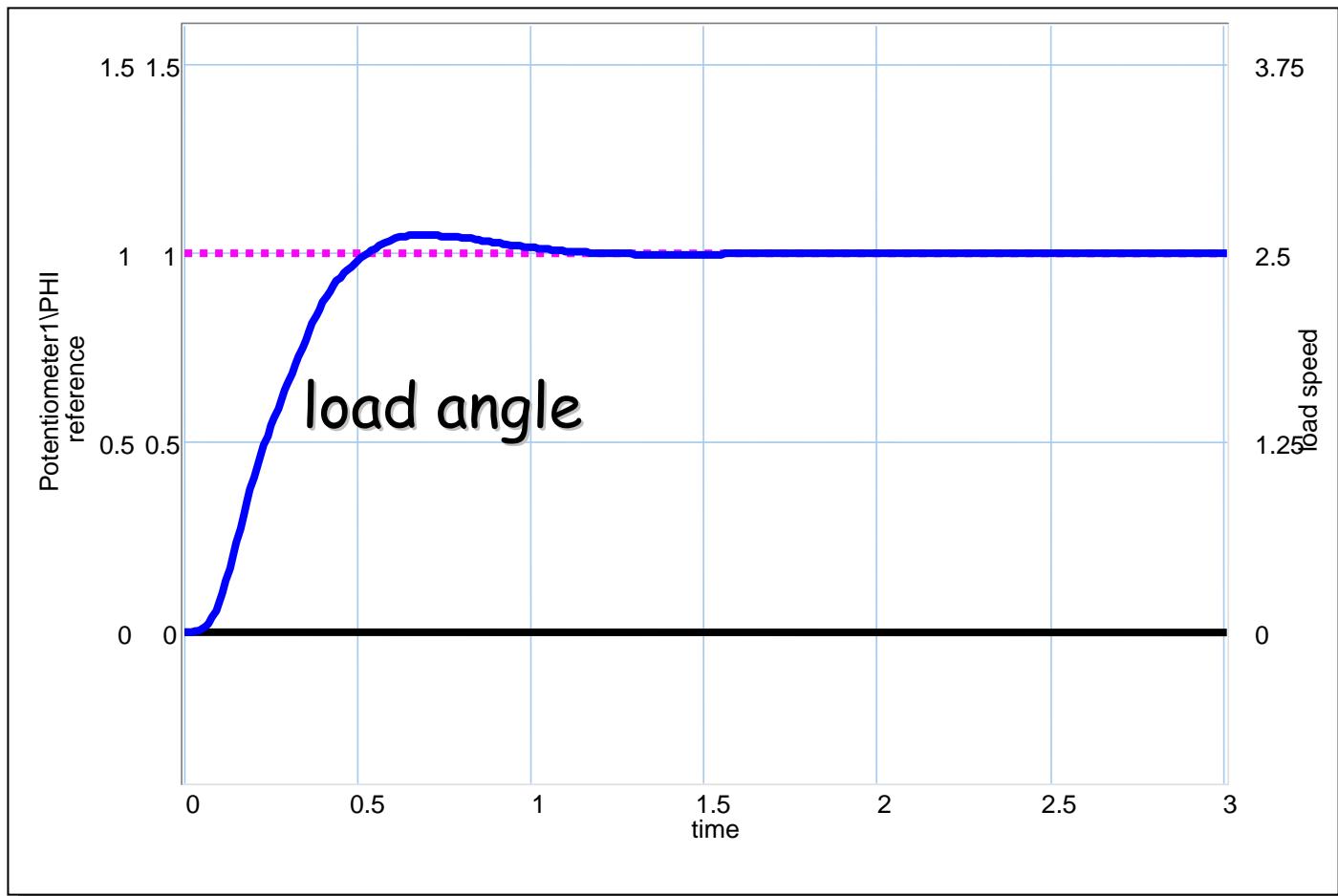
# Result in ‘real’ system



C\_Prop\_plus\_Tacho\_feedback.em

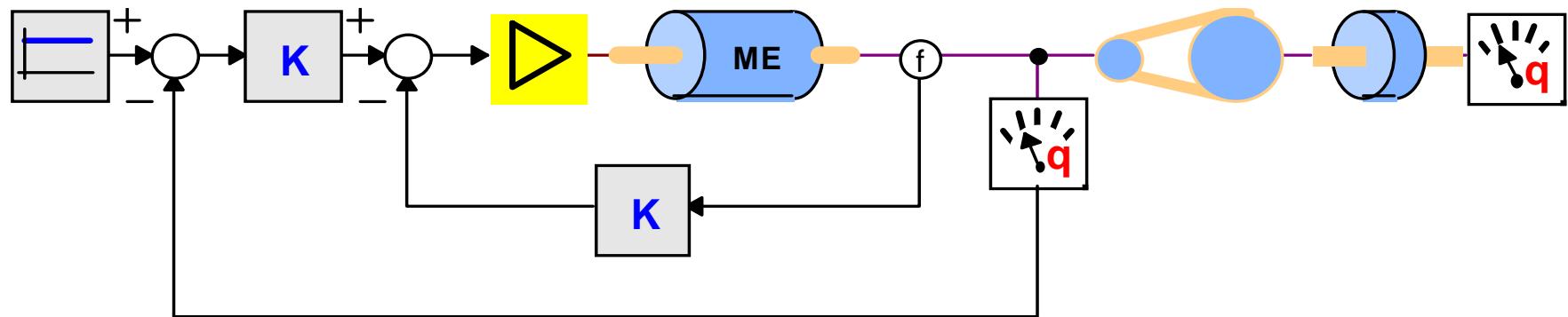
# Stepresponse

$$\begin{aligned}K_p &= 3 \\K_d &= 0.6/4 \\&= 0.15\end{aligned}$$



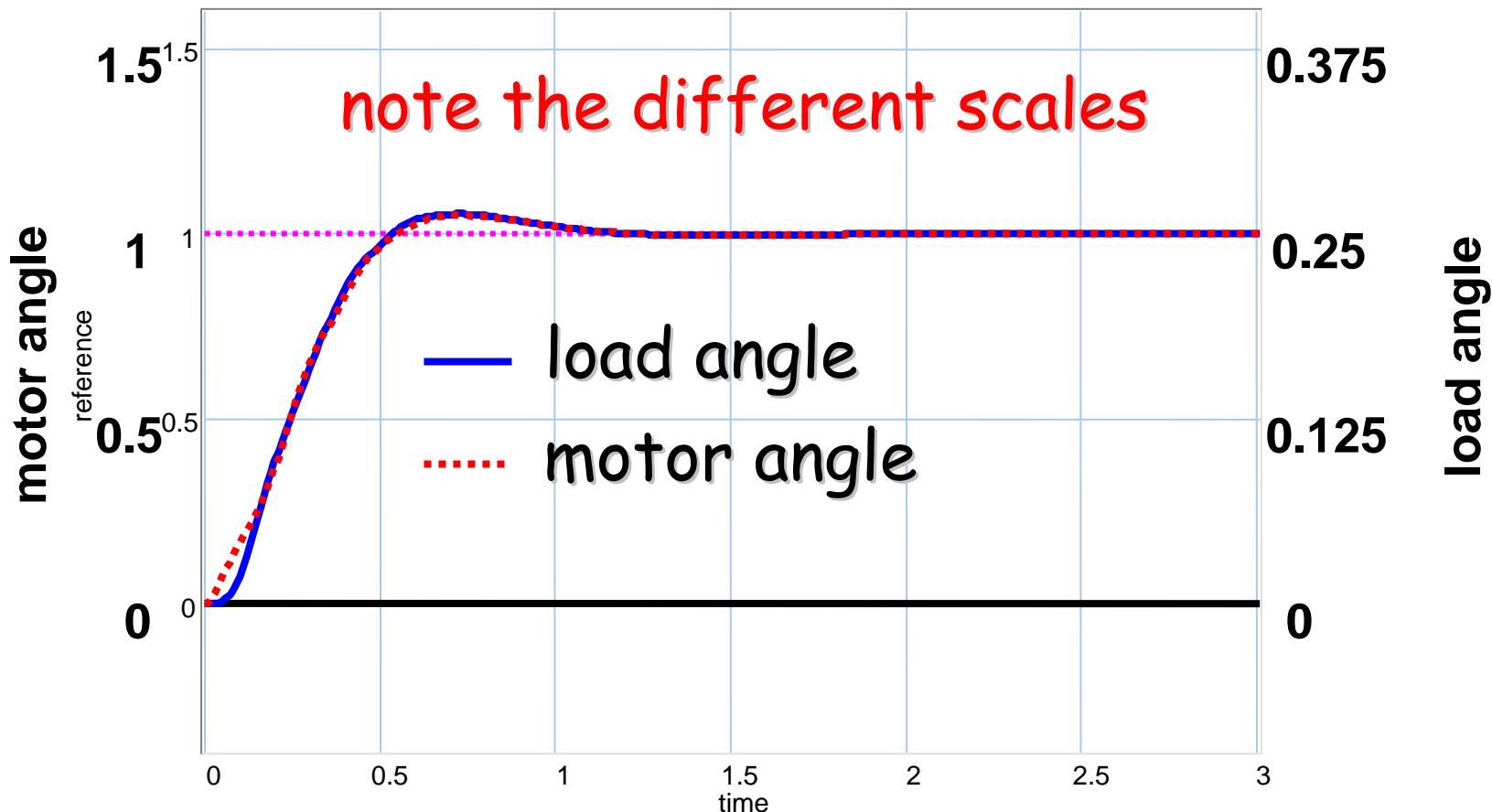
- Tacho feedback should always be based on the motor speed
- Tacho feedback of the load leads in a system with resonant poles easily to an unstable system

# Motor feedback only

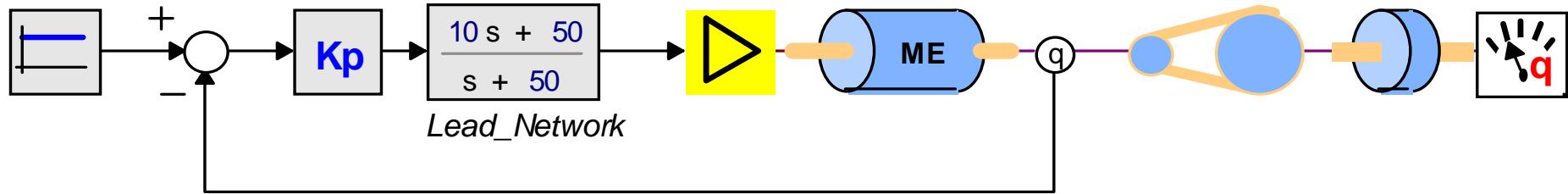


C1\_Prop\_plus\_Tacho\_feedback\_motor\_only.em

# Responses



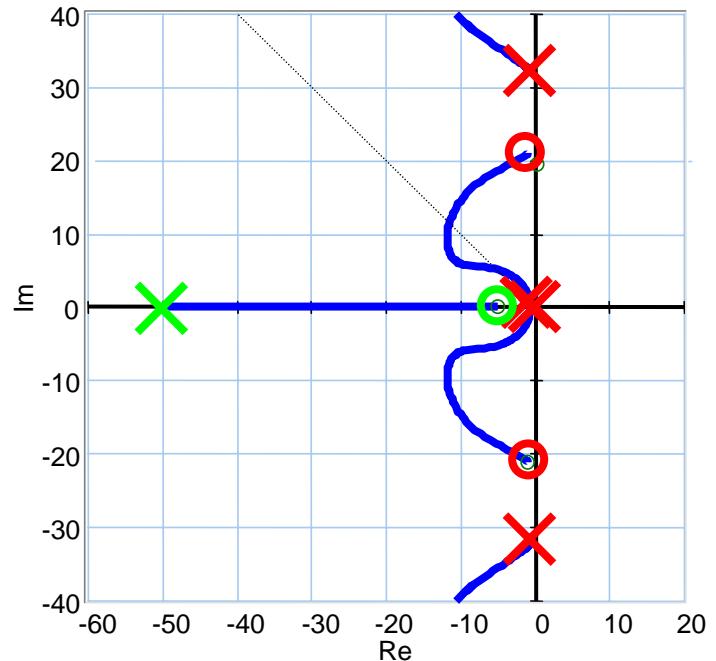
# Lead network

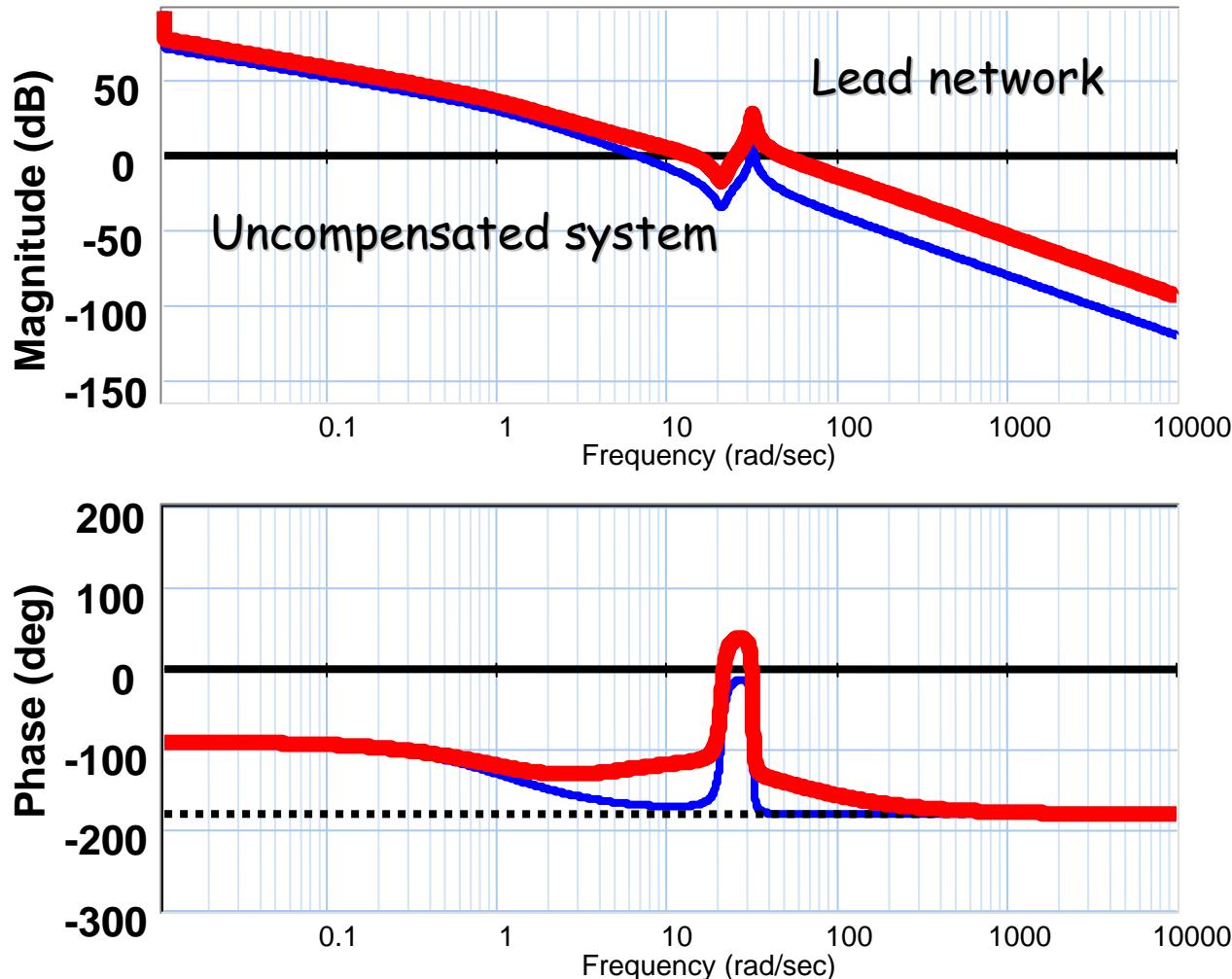


Lead network:

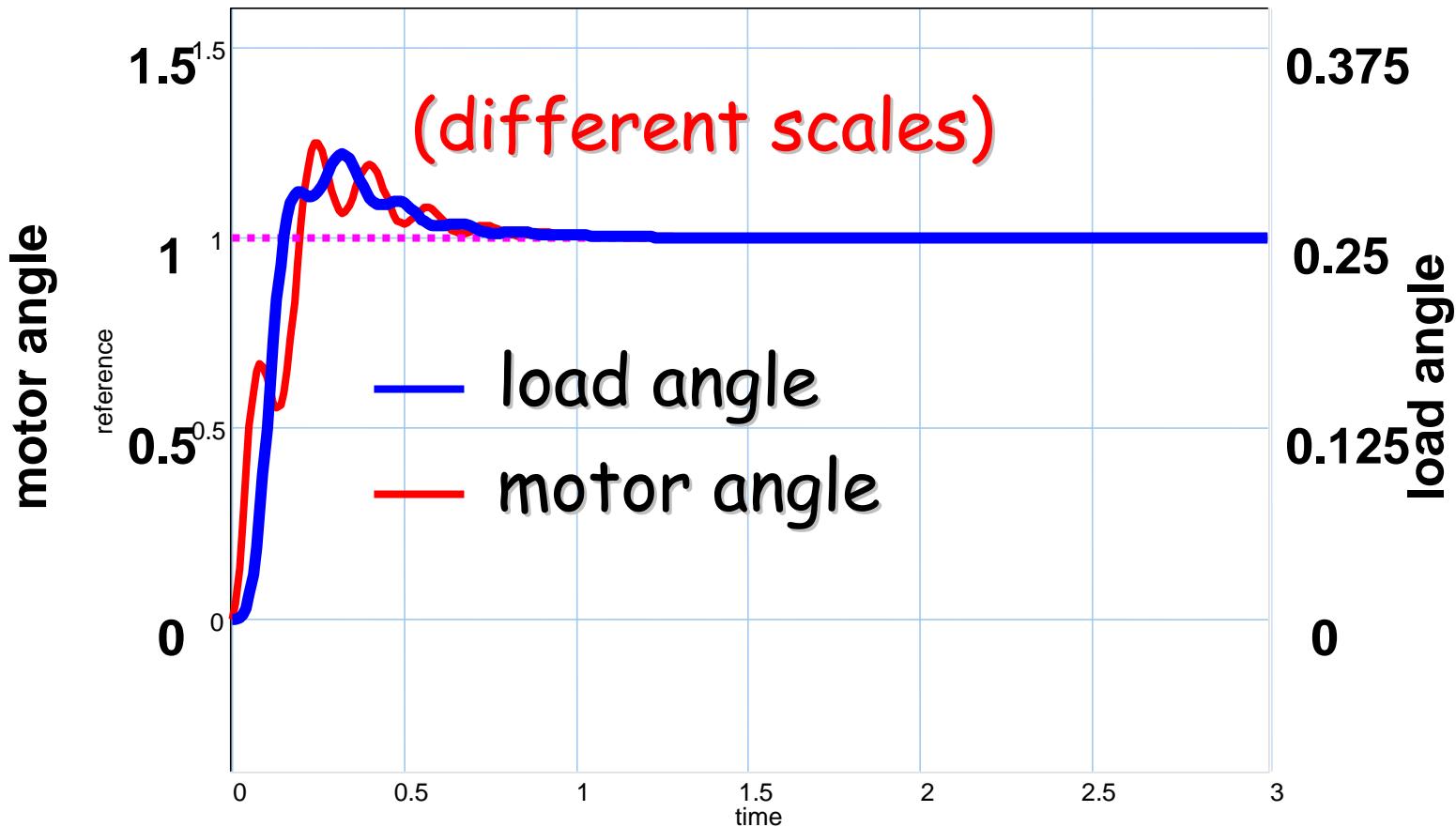
$K = 1.9$   
zero in  $-5$   
pole in  $-50$

C2\_lead\_network\_motor\_only\_no prefilter.em

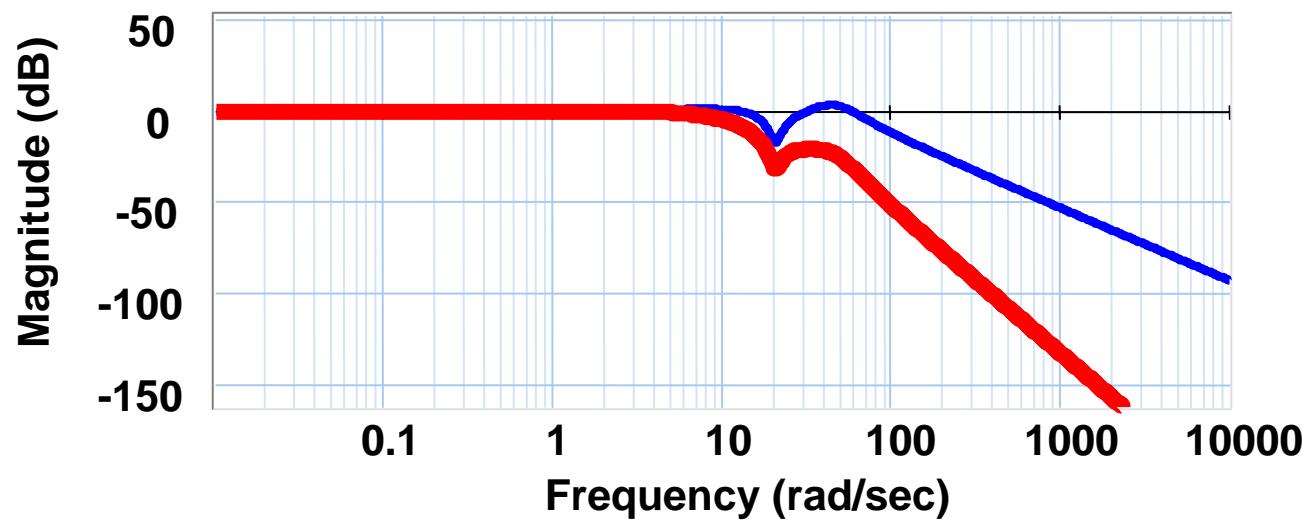




# Response

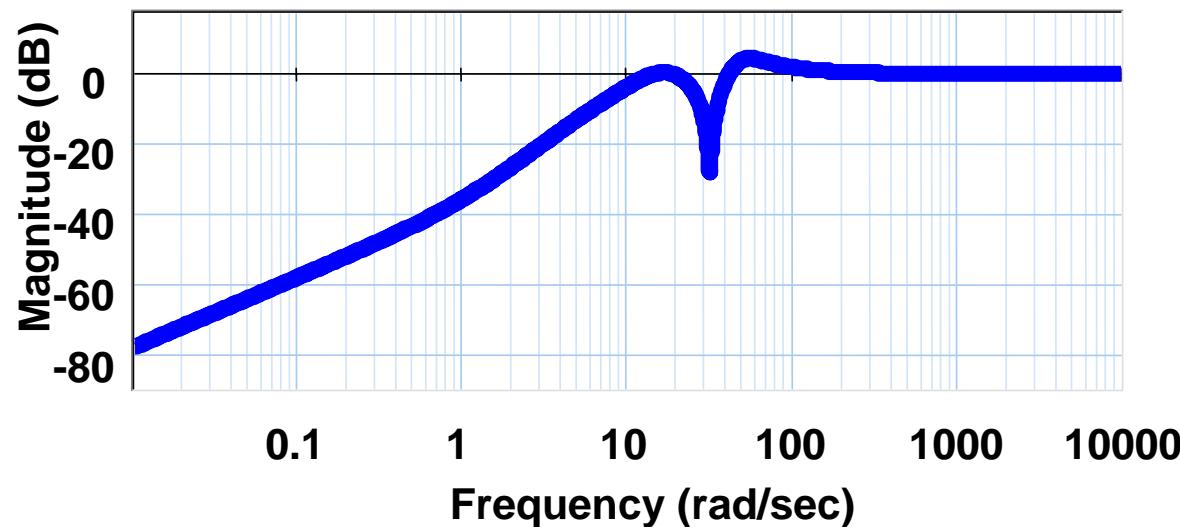


# Bode Closed-loop system

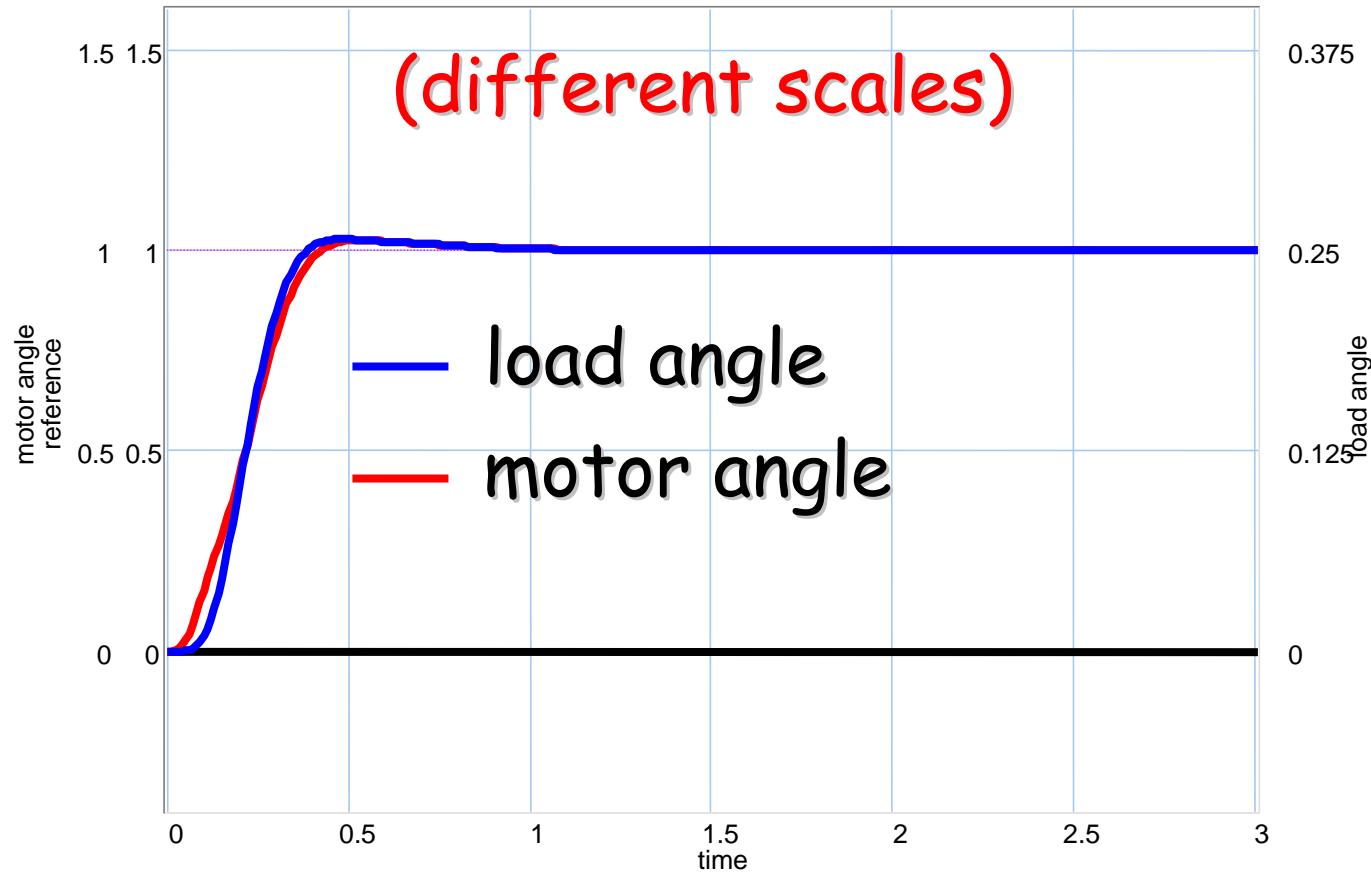


with prefilter in  $-10 \pm 2j$

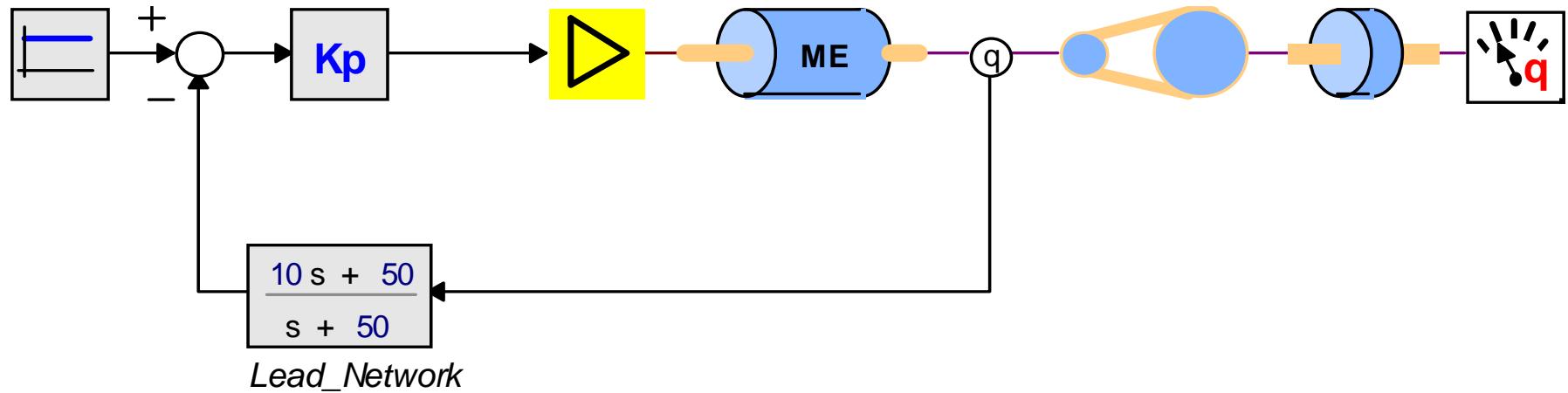
# Sensitivity



# Response with prefilter

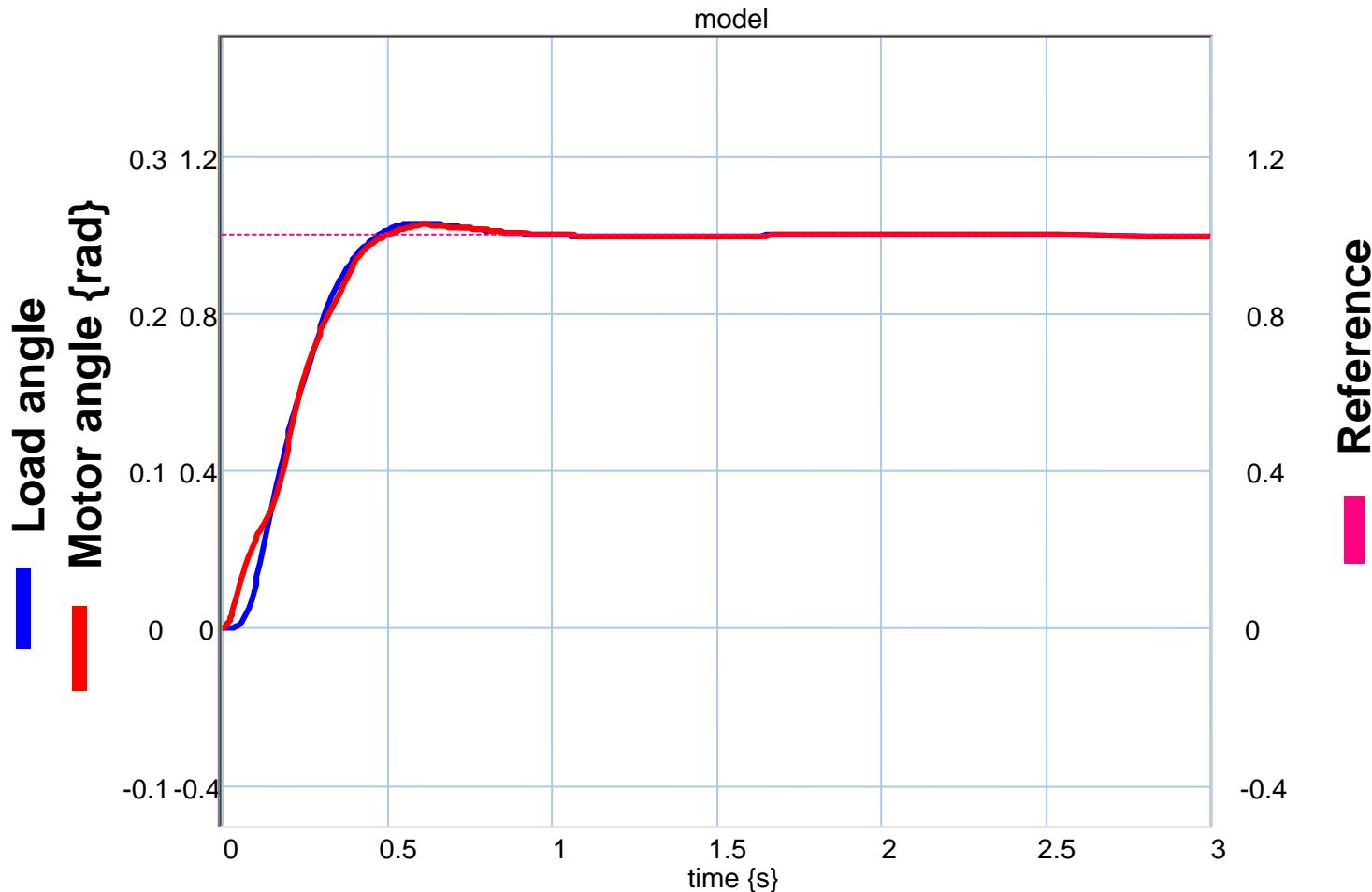


# Lead network in feedback

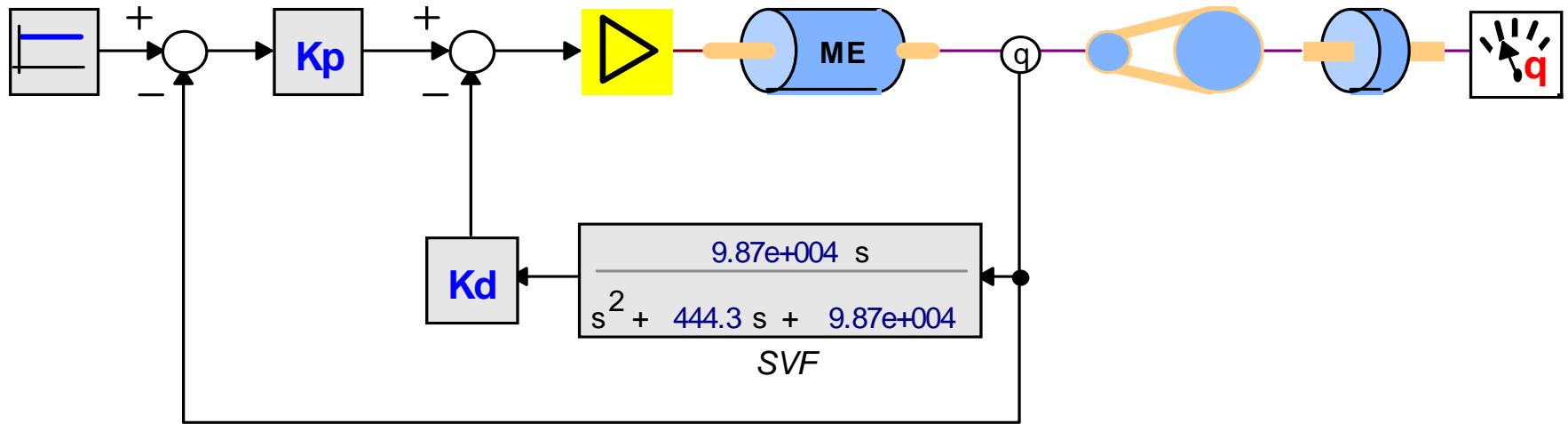


C2\_lead\_network\_in\_FB\_motor\_only.em

# Responses



# Motor angle feedback & derivative of motor angle

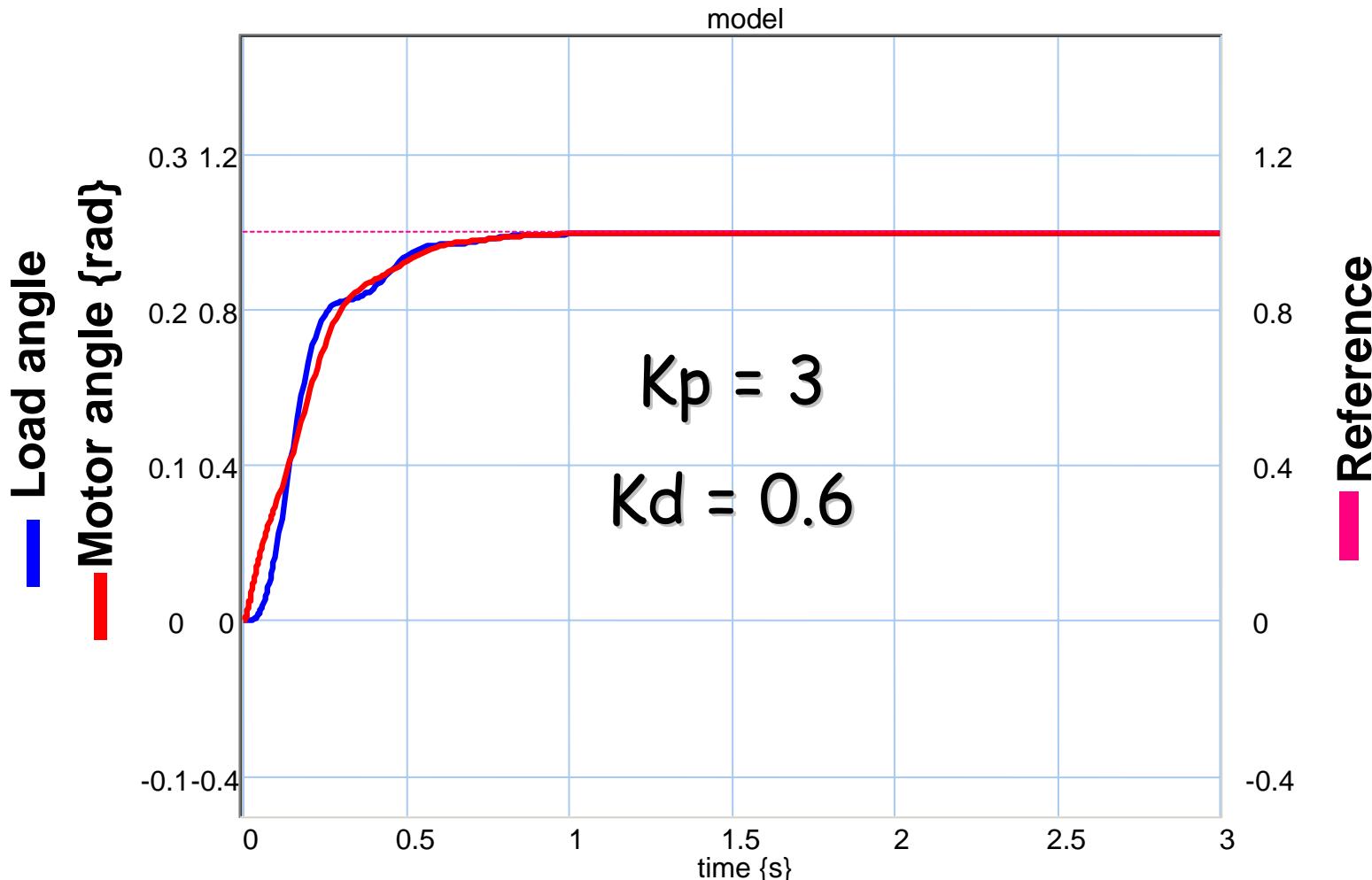


$$SVF: \frac{s\omega_n^2}{s^2 + 2\zeta\omega_n + \omega_n^2}$$

E1\_fb\_motor\_only\_SVF.em

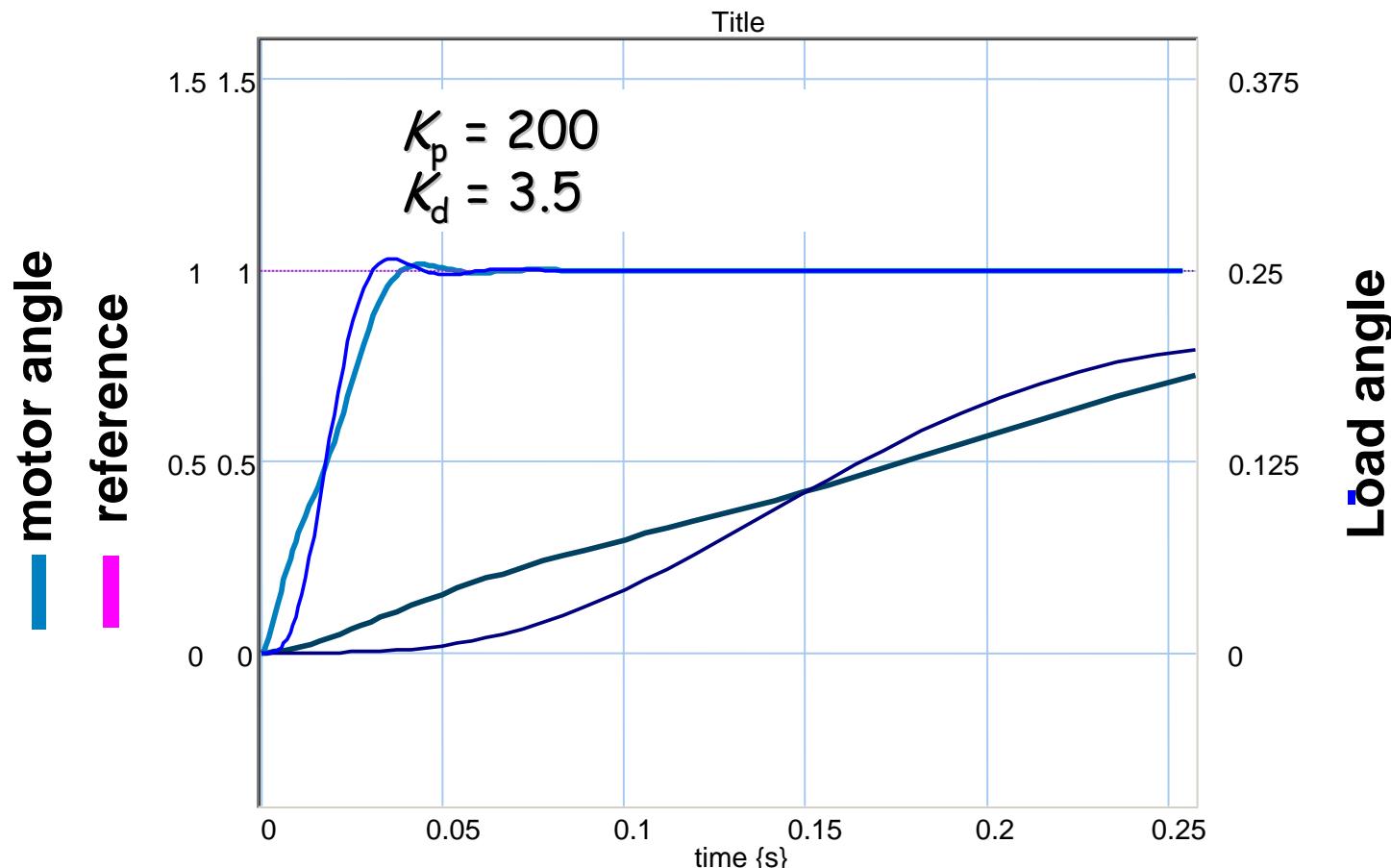
$\omega_n = 10$  times  $\omega$  resonant poles

# Responses



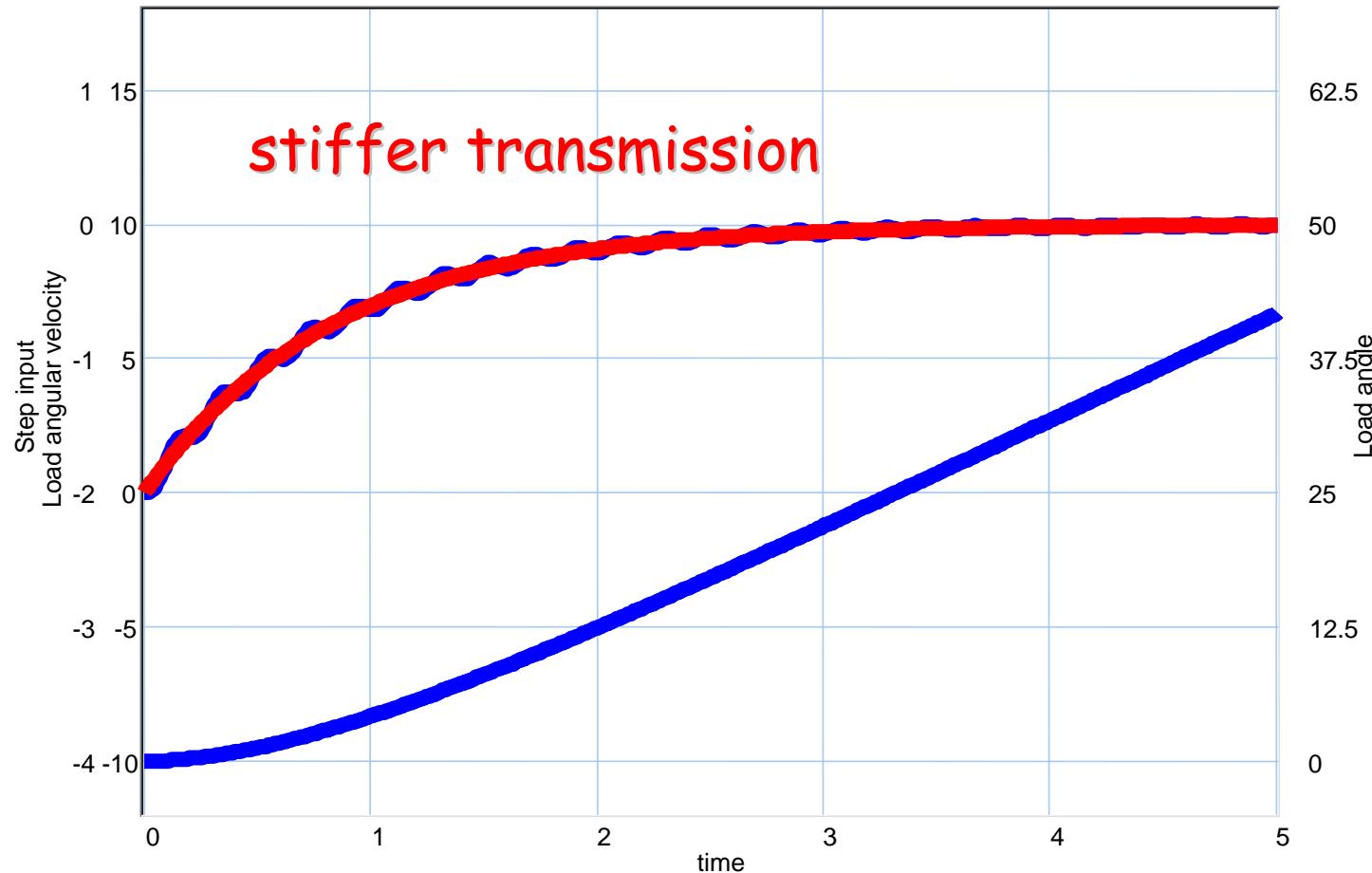
- Further improvements only possible by removing the limiting factor:
  - complex poles, due to limited stiffness of axis
  - a stiffer axis will move these poles to higher frequencies and
  - enable much better performance:
  - change  $k_{\text{axis}}$  from 1.45 into 100

# Response with a stiffer axis

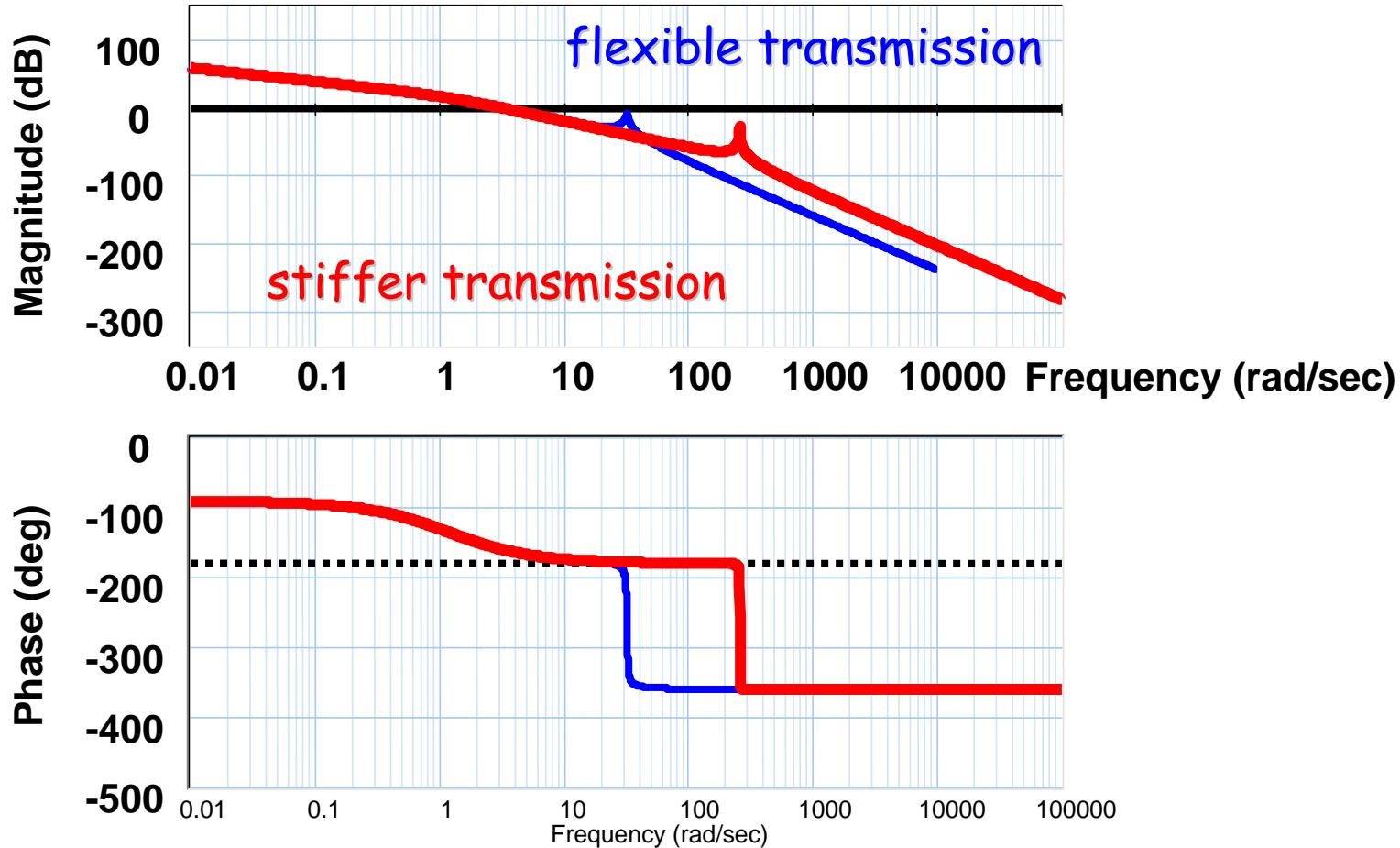


C\_Prop\_plus\_Tacho\_feedback\_stiff\_axis

# Open loop comparison

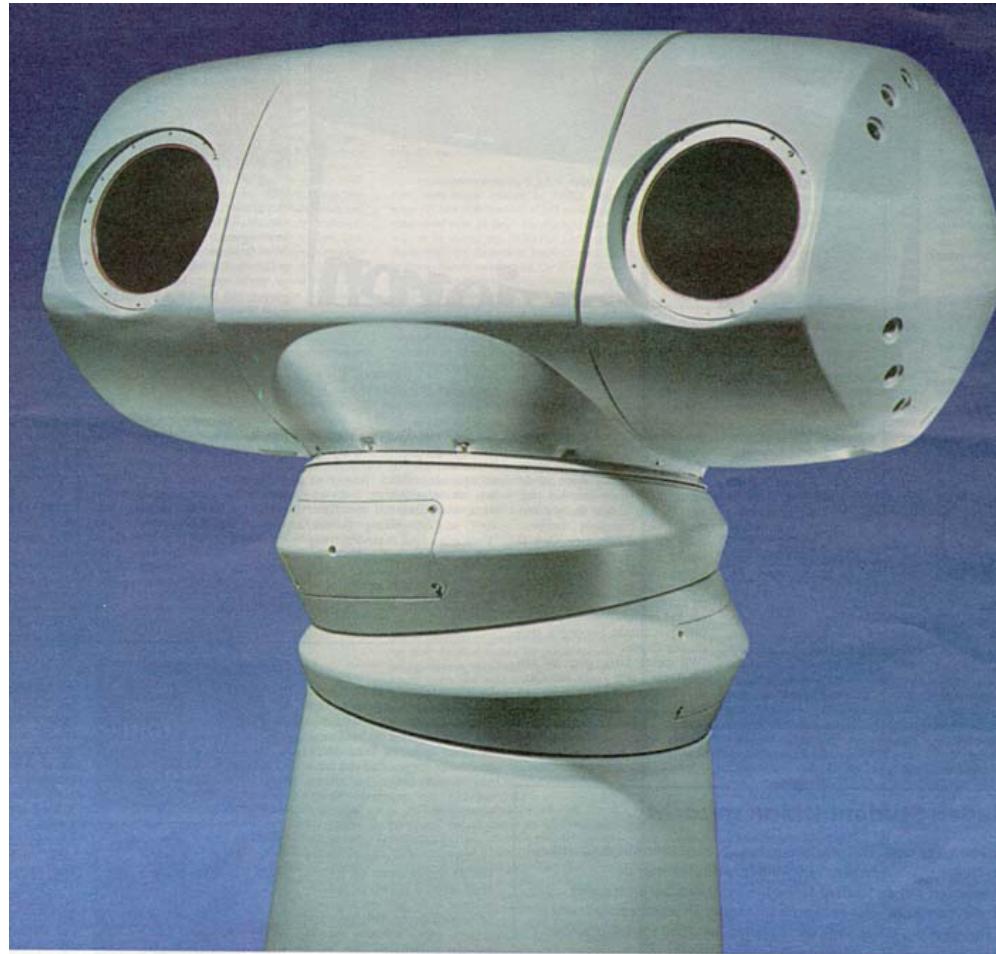


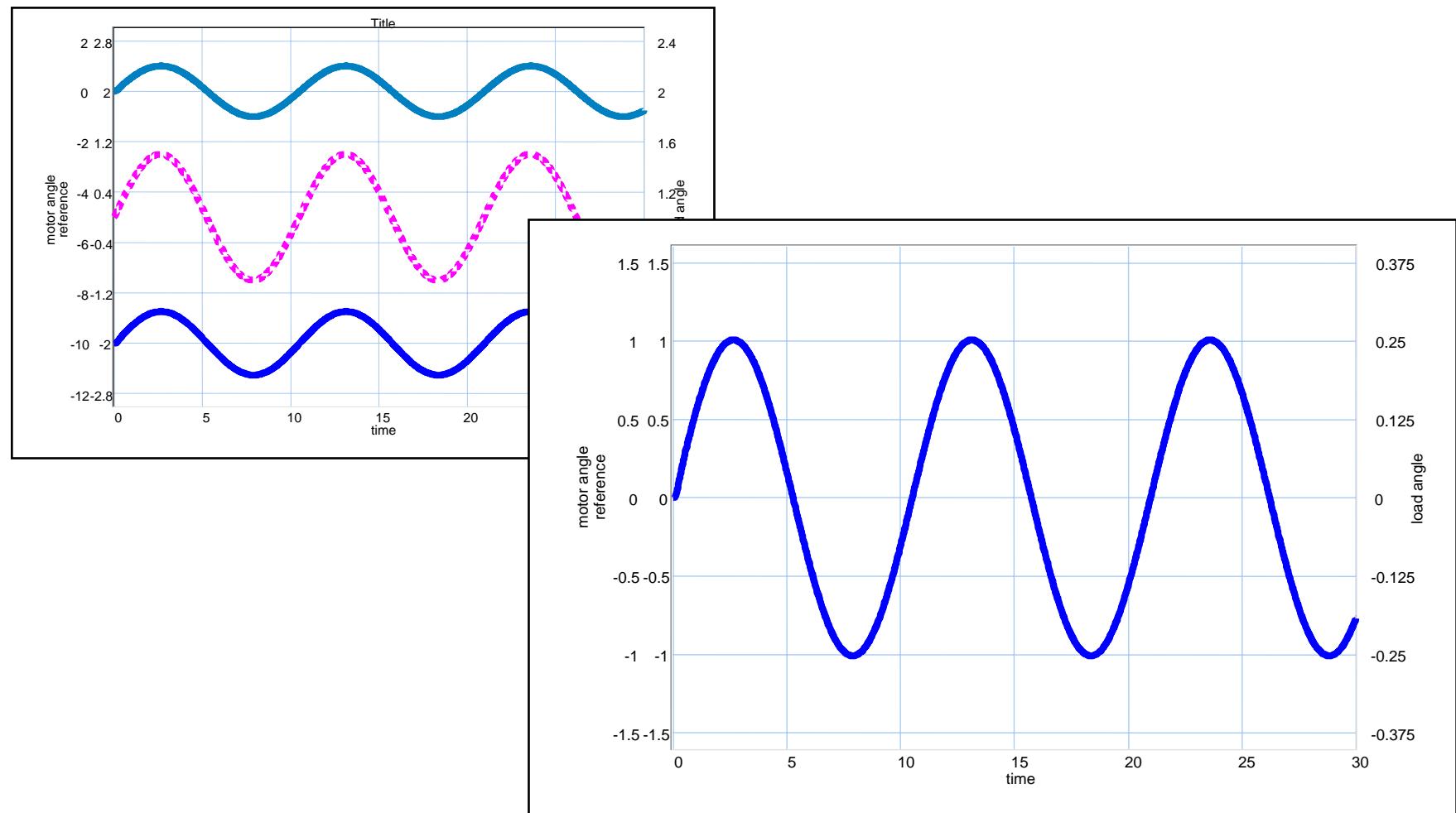
# Bode plots



# Tracking device

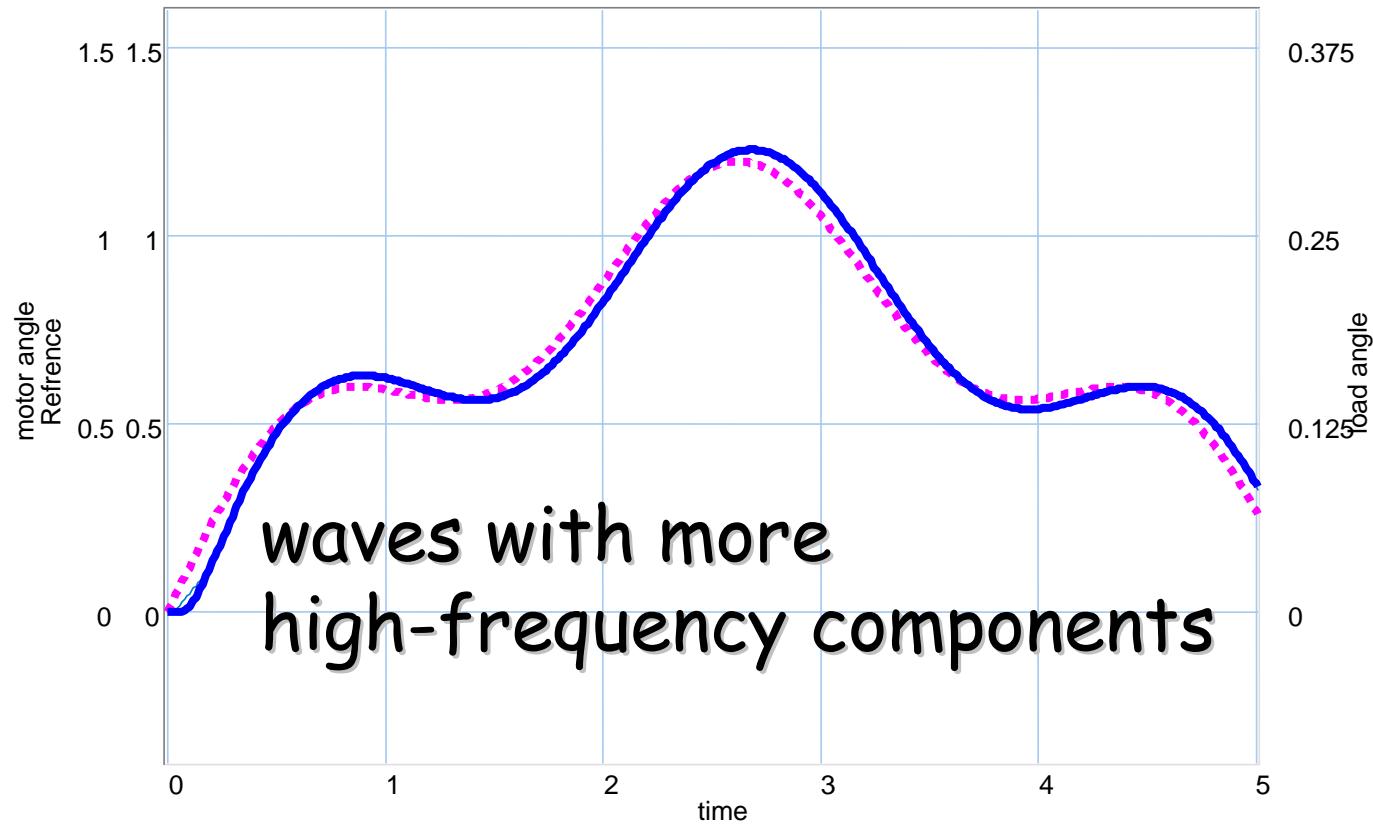
frequencies  
sea waves  
 $0.1 \text{ Hz}$   
 $0.6 \text{ rad/s}$



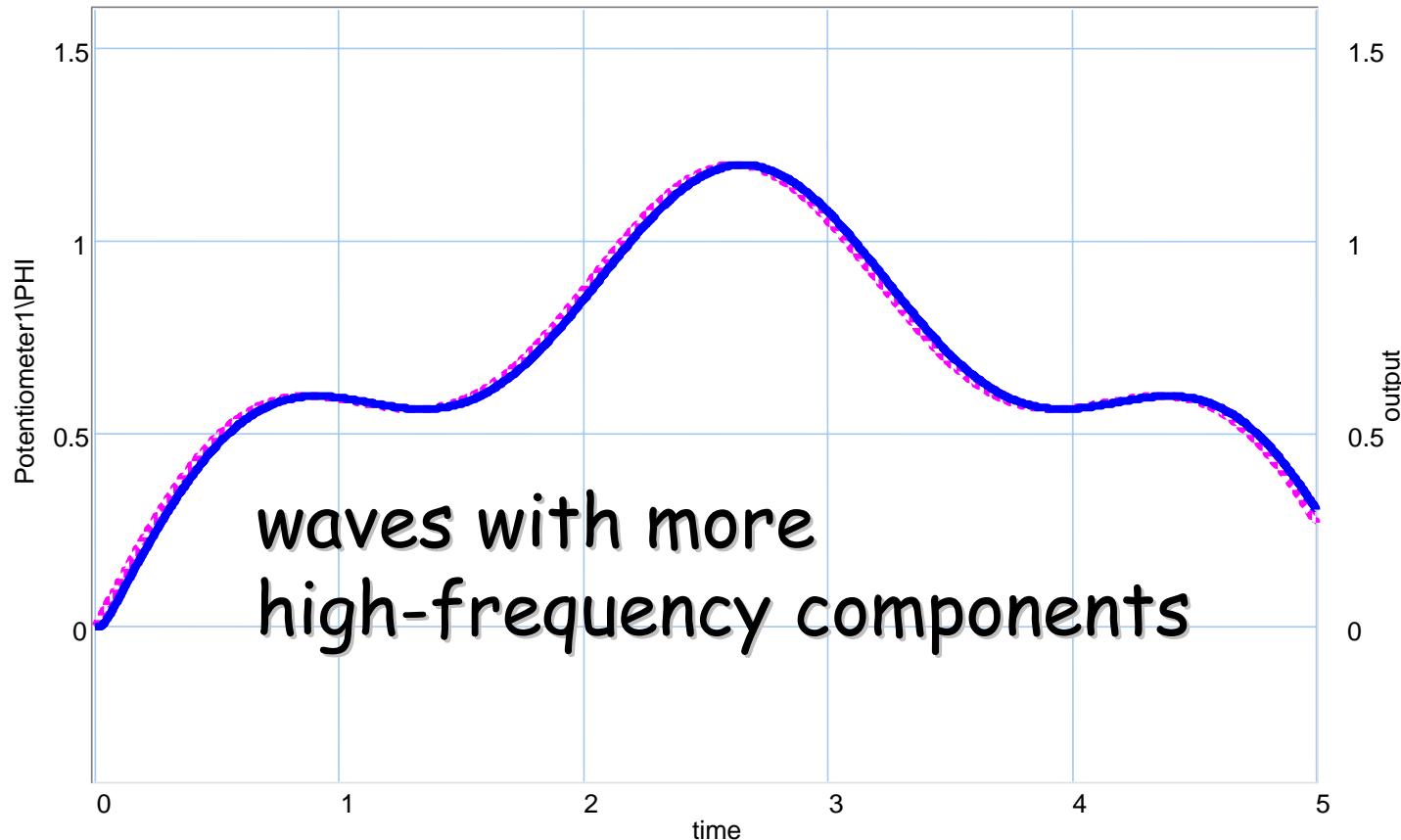


# Bandwidth too small

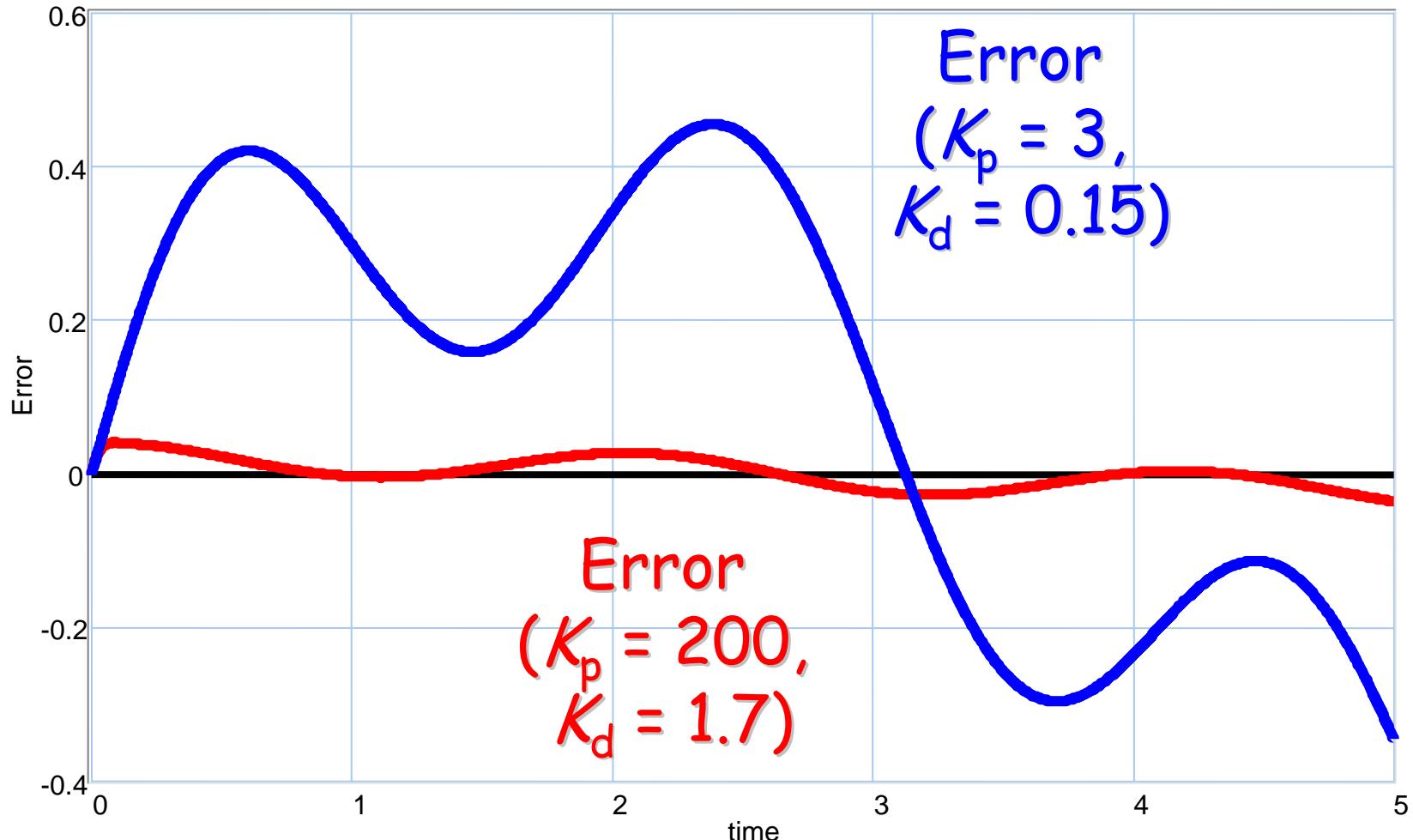
flexible transmission



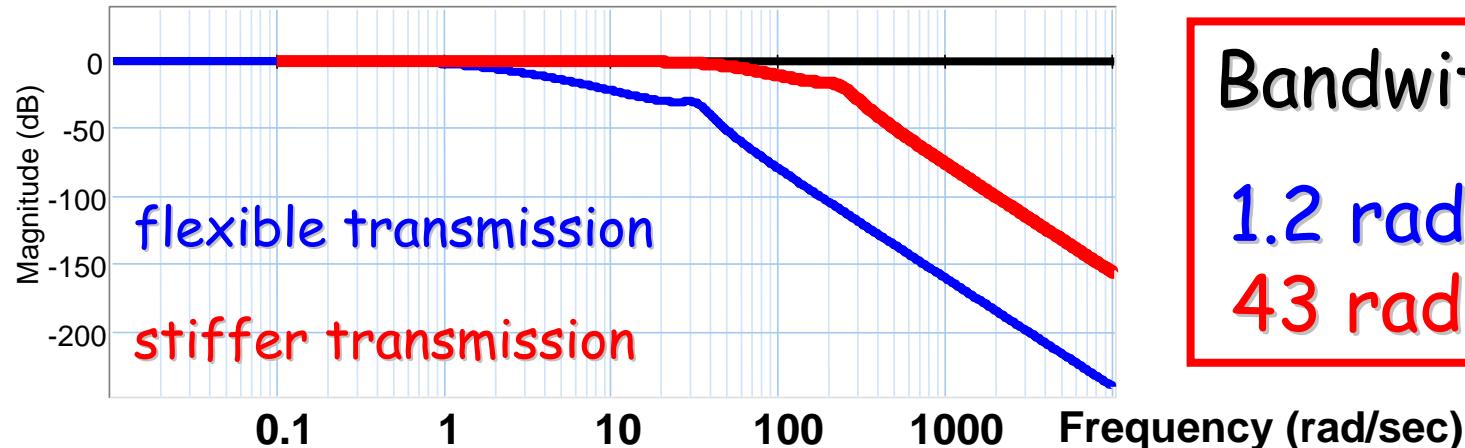
# Stiffer transmission



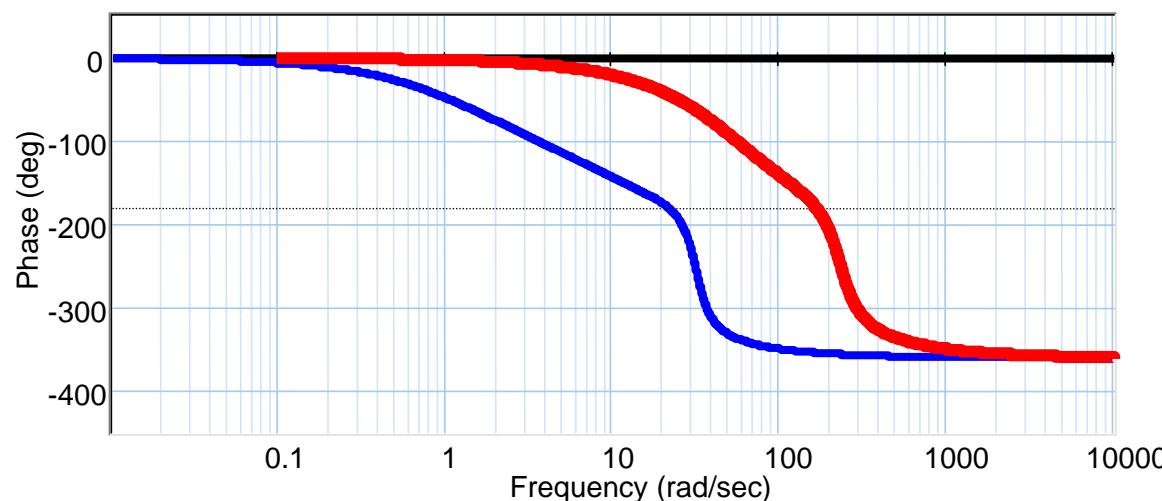
# Comparison



# Closed-loop Bode plots



Bandwidth:  
1.2 rad/s  
43 rad/s



- A realistic system can be controlled with the tools learned in this course
- What is left is the digital implementation of controllers:
  - Next two lectures